Active Versus Passive: Receiver Model Transforms for Diffusive Molecular Communication

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Abstract—This paper presents an analytical comparison of the active and passive receiver models in diffusive molecular communication. In the active model, molecules are absorbed when they collide with the receiver surface. In the passive model, the receiver is a virtual boundary that does not affect molecule behavior. Two approaches are presented to derive transforms between the active and passive receiver signals. As an example, we unify the two models for an unbounded diffusion-only molecular communication system with a spherical receiver. As time increases in the three-dimensional system, the transform functions have constant scaling factors, such that the receiver models are effectively equivalent. Methods are presented to enable the transformation of stochastic simulations, which are used to verify the transforms and demonstrate that transforming the simulation of a passive receiver can be more efficient and more accurate than the direct simulation of an absorbing receiver.

I. INTRODUCTION

Molecular communication has been receiving increasing attention as a strategy for the design of novel communication systems in fluid environments; see [1]. Much of this attention has been focused on molecular communication via diffusion, whose attractive properties include no required external energy or infrastructure for the molecules to propagate from a transmitter. Once molecules are released by the transmitter, they collide with other molecules in the fluid and move via a process that is effectively random.

Arguably one of the biggest schisms in the analysis of communication via diffusion is in the modeling of the receiver. Receiver models are generally classified as either passive or active. A passive receiver can observe but has no effect on molecule behavior. An active receiver is typically a site for chemical reactions, either inside or on its surface, and molecules can be identified when a reaction takes place. Passive receiver models are commonly favored for their simplicity in analysis and simulation, whereas active models are more realistic in representing the chemical detection of molecules.

The vast majority of the existing literature assumes one receiver model and then proceeds without assessing the choice. For example, active receivers have been considered in [2], [3], and passive receivers have been considered in [4] and the first author’s work in [5]. Explicit comparisons between the two models are very uncommon; one example is [6], where the one-dimensional passive and absorbing models were both fitted to experimental data obtained from the tabletop testbed in [7]. The authors observed similarities between the two models and attributed them to the dominance of airflow. Other authors, such as in [8], [9], derived results that were valid for either model but did not compare the models directly.

In this work, we compare the active and passive receiver models. Our goal is to unify the models so that we can transform signals from one model to the other. As an example, we unify the channel impulse responses (CIRs) for a spherical receiver in an unbounded diffusion-only environment. Other realistic phenomena, such as fluid flow or the potential for other chemical reactions, will be considered in future work.

We consider both a perfectly-absorbing receiver and a passive receiver in three dimensions (3D) and in one dimension (1D, i.e., the receiver is a line segment). All of these systems have been studied extensively, and we omit the two-dimensional environment because the CIR of the absorbing circle has not been described in closed form; see [1]. By carefully choosing how to represent the CIRs, we derive functions that transform a signal at a passive receiver into a signal at the corresponding absorbing receiver. In particular, we make the following contributions:

1) We take the integral of the passive CIR so that it includes prior signal information and is then comparable to the active CIR. In practice, this can be measured using the weighted sum detector that we introduced in [5]. Alternatively, we take the derivative of the active CIR so that it is an instantaneous measurement to compare with the passive CIR. The derivative can be obtained from the hitting rate detector proposed in [10]. These two approaches, which lead us to derive transform functions for the 1D and 3D environments, apply to absorbing and passive receivers in any environment.

2) We describe the proper implementation of the energy detector at a passive receiver and the hitting rate detector at an absorbing receiver to perform the integral and derivative operations, respectively. These detector modifications are needed to accurately apply the transforms between active and passive signals in simulations.

3) We demonstrate the accuracy of the transforms and their inverses, both analytically and via simulation. As a result, we can work with whichever model is currently most appropriate and then accurately transform the signal to the other model.

By deriving transforms between the passive and the
perfectly-absorbing receiver models for the sphere in a diffusion-only system, we effectively unify all existing literature that has selected one of these models. This unification makes the initial selection of a receiver model less critical. We can analyze one model and then simulate the other model, or divide the analysis between the two models, as desired, which gives us greater flexibility when deciding which model to use. As an example, we show that it can be both more accurate and more efficient to simulate an absorbing receiver signal by transforming a passive receiver simulation instead of directly simulating the absorbing receiver. This is due to the latter’s requirement of a very small simulation time step for accuracy.

We also note that our analysis focuses on expected CIRs. The impact of the transform functions on the channel statistics will be considered in future work. However, we verify the transforms numerically and also using average results obtained from our particle-based stochastic molecular communications simulator AcCoRD (Actor-based Communication via Reaction-Diffusion), which is available as a public beta [11].

The rest of this paper is organized as follows. Section II describes the 1D and 3D system models and presents the corresponding CIRs. In Section III, we derive the transform functions that describe the absorbing signal as a function of the passive signal. In Section IV, we discuss the implementation of the integral and derivative operations. We verify our analysis with numerical results and simulations in Section V, and conclude in Section VI.

II. SYSTEM MODEL AND CHANNEL IMPULSE RESPONSES

We consider a point transmitter (TX) releasing \( N \) molecules into an unbounded 1D or 3D environment. Molecules diffuse with constant diffusion coefficient \( D \), i.e., we assume uniform temperature and viscosity, and that the local molecule concentration is sufficiently low. These molecules are observed by a receiver (RX) that is centered at a distance \( d \) from the TX and has radius \( r_{RX} \), i.e., a segment of length \( 2r_{RX} \) in 1D and a sphere in 3D. If the RX is active, then we consider a perfectly-absorbing surface that removes and counts molecules as they arrive, i.e., the absorption rate is \( k \to \infty \). If the RX is passive, then it has no impact on molecule behavior but is able to count the number of molecules within its virtual boundary at any instant. Finally, we assume that diffusion is the only phenomenon affecting molecule behavior in the propagation environment (except for absorption at the active RX’s surface).

We define the channel impulse response (CIR) \( \overline{N}_{RX}(t) \) as the number of molecules expected at the RX at time \( t \), given that \( N \) molecules are instantaneously released by the TX at time \( t = 0 \). In the remainder of this section, we present the CIRs that are used throughout the remainder of this paper.

A. 3D Channel Impulse Responses

If the environment is 3D, then the CIR of the absorbing RX is given by [12, Eq. (23)]

\[
\overline{N}_{RX}(t)_{|3D}^{AB} = \frac{N_{RX}}{d} \text{erfc} \left( \frac{d - r_{RX}}{\sqrt{4Dt}} \right), \tag{1}
\]

where \( \text{erfc}(x) = 1 - \text{erf}(x) \) is the complementary error function from [13, Eq. (8.250.4)], and we use the “AB” superscript for “absorbing”. Eq. (1) describes the total number of molecules that have been absorbed by time \( t \). For the passive RX, the expected point concentration \( \overline{C}_{point}(t) \) is [14, Eq. (4.28)]

\[
\overline{C}_{point}(t)_{|3D}^{PA} = \frac{N}{(4\pi Dt)^{3/2}} \exp \left( -\frac{d^2}{4Dt} \right), \tag{2}
\]

where we use the “PA” superscript to denote “passive”.

It is common to assume that the molecule concentration is always uniform inside a passive RX, which is justified if the RX is sufficiently far from the TX, i.e., if \( d \gg r_{RX} \) (as we demonstrated in [15]). Here, we make this assumption for ease of analysis and clarity of presentation. Thus, we can multiply (2) by the RX volume \( V_{RX} \) and write the number of molecules expected inside the passive RX as

\[
\overline{N}_{RX}(t)_{|3D}^{PA} = \frac{NV_{RX}}{(4\pi Dt)^{3/2}} \exp \left( -\frac{d^2}{4Dt} \right). \tag{3}
\]

B. 1D Channel Impulse Responses

If the environment is 1D, then the CIR of the absorbing RX is given by [1, Eq. (7)]

\[
\overline{N}_{RX}(t)_{|1D}^{AB} = N_{RX} \text{erfc} \left( \frac{d - r_{RX}}{\sqrt{4Dt}} \right). \tag{4}
\]

The expected point concentration for the 1D passive RX is [14, Eq. (3.6)]

\[
\overline{C}_{point}(t)_{|1D}^{PA} = \frac{N}{\sqrt{4\pi Dt}} \exp \left( -\frac{d^2}{4Dt} \right). \tag{5}
\]

If we assume that the passive RX is sufficiently far from the TX, then the simplified 1D CIR is directly from (5) as

\[
\overline{N}_{RX}(t)_{|1D}^{PA} = \frac{r_{RX}N}{\sqrt{\pi Dt}} \exp \left( -\frac{d^2}{4Dt} \right). \tag{6}
\]

III. UNIFYING RECEIVER MODELS

In this section, we seek the existence of transform functions that have the form

\[
\text{Absorbing Signal} \overset{?}{=} S(\text{Passive Signal}), \tag{7}
\]

where we emphasize that we are most interested in obtaining an active RX signal from a passive RX, since a passive RX is generally easier to simulate (though the inverse transform \( S^{-1} \) is also of interest). We do not constrain the “signals” in (7) to be CIRs; rather, we seek to manipulate either the absorbing or the passive CIR so that it is comparable to the other.

We claim that the CIRs of the two receiver models can be perceived as similar measurements but they provide fundamentally different information about what has happened at the RX. A sample of the absorbing RX’s CIR is the total number of molecules that have arrived at that RX since they were released by the TX, i.e., a sample includes history information. However, a sample of the passive RX’s CIR is only the current number of molecules that are inside the RX at the instant when the sample is taken; the history of the molecules that have
entered and left the passive RX is ignored. So, to derive a transform, we propose using either a passive receiver model that accounts for signal “history”, or an active receiver model that describes the instantaneous behavior.

We first consider the 3D environment, where we derive the transform function that applies to the passive signal with history information, \( S'_{3D} \), and that applies to the instantaneous passive signal, \( S''_{3D} \). These transforms will simplify to constant scaling factors in the asymptotic case, i.e., as \( t \to \infty \). Then, we consider the 1D environment, where the corresponding transform functions are \( S'_{1D} \) and \( S''_{1D} \).

### A. 3D Analysis

We begin our 3D analysis with the more intuitive signal manipulation, which is to model the receiver as an energy detector. The CIR of the absorbing RX is intuitively a measure of the received energy over time, since the signal accounts for every molecule arrival. Analytically, an energy detector for the passive RX is defined by integrating the CIR over time, and we will show in Section IV that this can be implemented with a weighted sum detector. Analogously to [14, Eq. (3.5b)], and we will show in Section IV that this can be implemented in the passive RX is defined by integrating the CIR over time, since the signal accounts for signal “history”, or an active receiver model.

To write (8) as a function of (1), we need a way to separate the terms inside the complementary error function in (1). To do this, we will use the elementary approximation of \( \text{erf}(x) \) in (16), which has a relative error of less than 1% for all positive \( x \), to re-write \( \text{erf}(x) \) as

\[
\text{erf}(x) \approx \exp\left(\frac{-16}{23}x^2 - \frac{2}{\sqrt{\pi}}x\right),
\]

and therefore write \( \text{erf}(\cdot) \) in (1) as

\[
\text{erf}(\frac{d - r_{RX}}{\sqrt{4Dt}}) \approx \text{erf}(\frac{d}{\sqrt{4Dt}}) A(t),
\]

where we define the function \( A(t) \) as

\[
A(t) = \exp\left(\frac{r_{RX}}{\sqrt{Dt}} \left(\frac{8(2d - r_{RX})}{23\sqrt{4Dt}} + \frac{1}{\sqrt{\pi}}\right)\right).
\]

Using (10) and the equation for the volume of a sphere, we write the transform function \( S'_{3D} \) as

\[
\Delta N_{RX}(t) |_{3D}^{AB} = S'_{3D}(ED(t) |_{3D}^{PA}) \approx 3DA(t) \frac{r_{RX}}{r_{RX}^2} ED(t) |_{3D}^{PA},
\]

and we can re-arrange (12) to find the inverse transform \( S''_{3D}^{-1} \).

In the asymptotic case, i.e., as \( t \to \infty \), the complementary error function goes to 1 and thus the transform function \( S''_{3D} \) has a constant scaling factor

\[
\Delta N_{RX}(t) |_{3D}^{AB} = S''_{3D}(ED(t) |_{3D}^{PA}) \approx 3\frac{D}{r_{RX}} ED(t) |_{3D}^{PA}.
\]

Although arriving at (13) is straightforward, the impact is notable. We see that, asymptotically, these two fundamentally different receiver models are effectively equivalent.

Next, we perform the complementary signal manipulation, i.e., we seek a measure of the instantaneous behavior of the absorbing receiver model to compare with the passive RX CIR. Thus, we should consider taking the derivative of the active RX’s CIR. This result is the rate of molecule absorption at the RX, \( \Delta N_{RX}'(t) \), and has been previously presented in the molecular communication literature as [12, Eq. (22)]

\[
\Delta N_{RX}'(t) |_{3D}^{AB} = \frac{N_{RX}(d - r_{RX})}{d\sqrt{4\pi Dt^3}} \exp\left(-\frac{(d - r_{RX})^2}{4Dt}\right).
\]

The implementation of (14) in simulations will be discussed in Section IV. Here, we can compare (14) with (3) (once again with a slight abuse of notation since (14) is in \([\text{mol} \cdot \text{s}^{-1}]\) and (3) is in \([\text{mol}]\). From the properties of exponential functions, it can be shown that the transform function \( S''_{3D} \) has a constant scaling factor, i.e.,

\[
\Delta N_{RX}(t) |_{3D}^{AB} = S''_{3D}(ED(t) |_{3D}^{PA}) \approx 3\frac{D}{r_{RX}^2} ED(t) |_{3D}^{PA}.
\]

and if we again assume that the TX is sufficiently far from the RX (i.e., \( d \gg r_{RX} \)), then the two asymptotic transform functions \( S''_{3D} \) and \( S''_{3D}^{\infty} \) are analogous for the 3D receiver models with the same scaling factor, i.e., \( 3D/r_{RX}^2 \).

### B. 1D Analysis

Our strategy to transform between the active and passive receiver models in the 1D system is the same as that applied for the 3D system. The energy detector signal for the 1D system can be obtained by integrating the passive CIR in (6), i.e., we seek a measure of the instantaneous behavior of the RX. Thus, we should consider taking the derivative of the active receiver model to compare with the passive RX CIR. This result is the rate of molecule absorption at the RX, \( \Delta N_{RX}'(t) \), and has been previously presented in the molecular communication literature as [12, Eq. (22)]

\[
\Delta N_{RX}'(t) |_{1D}^{AB} = \frac{N_{RX}(d - r_{RX})}{d\sqrt{4\pi Dt^3}} \exp\left(-\frac{(d - r_{RX})^2}{4Dt}\right).
\]

The implementation of (14) in simulations will be discussed in Section IV. Here, we can compare (14) with (3) (once again with a slight abuse of notation since (14) is in \([\text{mol} \cdot \text{s}^{-1}]\) and (3) is in \([\text{mol}]\). From the properties of exponential functions, it can be shown that the transform function \( S''_{3D} \) has a constant scaling factor, i.e.,

\[
\Delta N_{RX}(t) |_{1D}^{AB} = S''_{3D}(ED(t) |_{1D}^{PA}) \approx 3\frac{D}{r_{RX}^2} ED(t) |_{1D}^{PA}.
\]

and if we again assume that the TX is sufficiently far from the RX (i.e., \( d \gg r_{RX} \)), then the two asymptotic transform functions \( S''_{1D} \) and \( S''_{1D}^{\infty} \) are analogous for the 1D receiver models with the same scaling factor, i.e., \( 3D/r_{RX}^2 \)
We compare (18) with the 1D absorbing CIR in (4). We do so to obtain the transform function \( S''_{1D} \) as
\[
\Delta N_{RX}(t)_{|AB}^{1D} = S''_{1D}(\mathbf{ED}(t)|^{PA}_{1D}) \\
= \frac{2DA(t)}{d} \left[ N \sqrt{\frac{t}{\pi D}} \exp\left(-\frac{d^2}{4DT}\right) - \mathbf{ED}(t)|^{PA}_{1D} \right],
\]
(19)
where \( A(t) \) is in (11). As \( t \to \infty \), the transform becomes
\[
\Delta \mathbf{N}_{RX}(t)_{|AB}^{1D} = S''_{1D}(\mathbf{ED}(t)|^{PA}_{1D}) \\
\approx \frac{2D}{d} \left[ N \sqrt{\frac{t}{\pi D}} \frac{r_{RX}}{2\pi r_{RX}} \right],
\]
(20)
which does not have a constant scaling factor.

Finally, we determine the transform function \( S''_{1D} \). The derivative of the CIR at the 1D absorbing RX in (4) is the rate of molecule absorption at that RX, and from [1, Eq. (6)] has the form
\[
\Delta N_{RX}(t)_{|AB}^{1D} = \frac{N(d - r_{RX})}{\sqrt{4\pi Dt}} \exp\left(-\frac{(d - r_{RX})^2}{4Dt}\right).
\]
(21)

We can compare (21) with (6) to derive the transform function \( S''_{1D} \) as
\[
\Delta \mathbf{N}_{RX}(t)_{|AB}^{1D} = S''_{1D}(\mathbf{N}_{RX}(t)|^{PA}_{1D}) \\
= \frac{d - r_{RX}}{2\pi r_{RX}} \mathbf{N}_{RX}(t)|^{PA}_{1D} \exp\left(\frac{r_{RX}(2d - r_{RX})}{4Dt}\right),
\]
(22)
which closely resembles its 3D variant \( S''_{3D} \) in (15). However, one key difference is that it does not have a constant scaling factor in the asymptotic case, i.e.,
\[
\Delta \mathbf{N}_{RX}(t)_{|AB}^{1D} = S''_{1D}(\mathbf{N}_{RX}(t)|^{PA}_{1D}) \approx \frac{d - r_{RX}}{2\pi r_{RX}} \mathbf{N}_{RX}(t)|^{PA}_{1D}.
\]
(23)

The 1D and 3D environments that we considered have very similar configurations but resulted in different transform functions to convert between the active and passive receiver models. For reference, we have summarized the equations for the signals and their transforms in Table I. We note that the effective equivalence (via the constant scaling factor \( 3D/r_{RX}^2 \)) that we observed for the 3D model in the asymptotic case was not observed for the 1D model. Even though each transform function only applies to its corresponding system model, our approach to derive the transform functions could be applied to any system model where comparable CIRs can be found.

### IV. Implementation Details

In this section, we briefly describe how to measure the energy and absorption rate detector signals so that we can verify the transforms for active and passive receivers via simulations. To do so, we need to accommodate discrete (and noisy) observations of the CIRs and not just the analytical expressions in Section II. We assume that, for either receiver model, the RX takes \( M \) samples between time \( t = 0 \) and time \( t = T \). The samples are equally spaced in time by sampling interval \( \Delta t_M = T/M \). The individual sample taken at time \( t_m \) is labeled as \( N_{RX}(t_m) \), where \( t_m = m\Delta t_M, m \in \{1, 2, \ldots, M\} \).

First, we consider the energy detector. Energy detection by a passive RX has been considered analytically in [4], [17] as the integration of the continuous CIR (e.g., (8)). To accommodate discrete observations, we consider the weighted sum detector that we introduced in [5, Eq. (37)], where the weighted sum is
\[
\sum_{m=1}^{M} w_m N_{RX}(t_m),
\]
(24)
and \( w_m \) is the \( m \)th weight. In [5], we described the equal weight detector, where \( w_m = 1 \forall m \), as analogous to an energy detector. However, the corresponding sum is not precisely an energy detector. To be an energy detector, each sample must also be scaled by the sampling interval \( \Delta t_M \) (as done in [18]). If not, then we can artificially detect more energy if we take more samples over the same sampling interval. By including the sampling interval, we design the discrete energy detector for the passive RX ED \( \cdot \) as
\[
\mathbf{ED}\left( t_m + \frac{\Delta t_M}{2} \right)|^{PA}_{1D} = \Delta t_M \sum_{l=1}^{M} N_{RX}(t_l),
\]
(25)
which applies to any environment, and we add the shift \( \frac{\Delta t_M}{2} \) to the energy detector observation time because the \( l \)th sample accounts for the energy in the continuous interval \([t_l - \frac{\Delta t_M}{2}, t_l + \frac{\Delta t_M}{2}]\).

The net number of absorbed molecules is used in [10] as an absorption rate detector, but the number of molecules arriving over the interval \([t_m, t_{m+1}]\), i.e.,
\[
N_{RX}(t_{m+1}) - N_{RX}(t_m),
\]
(26)
is only a legitimate rate if we divide the difference by the sampling interval. Therefore, to be consistent with the absorption
TABLE II
SIMULATION SYSTEM PARAMETERS.

<table>
<thead>
<tr>
<th>Parameter</th>
<th>Symbol</th>
<th>Units</th>
<th>System 1</th>
<th>System 2</th>
</tr>
</thead>
<tbody>
<tr>
<td># of Dimensions</td>
<td>-</td>
<td>-</td>
<td>3</td>
<td>1</td>
</tr>
<tr>
<td>RX Radius</td>
<td>rRX</td>
<td>µm</td>
<td>1</td>
<td>0.5</td>
</tr>
<tr>
<td>Molecules Released</td>
<td>N</td>
<td>mol</td>
<td>10^4</td>
<td>10^3</td>
</tr>
<tr>
<td>Distance to RX</td>
<td>d</td>
<td>µm</td>
<td>5</td>
<td>5</td>
</tr>
<tr>
<td>Diffusion Coeff.</td>
<td>D</td>
<td>m^2/s</td>
<td>10^-9</td>
<td>10^-9</td>
</tr>
<tr>
<td>Sampling Period</td>
<td>∆t_M</td>
<td>µs</td>
<td>40</td>
<td>2 x 10^3</td>
</tr>
<tr>
<td>Passive Time Step</td>
<td>∆t_sim</td>
<td>µs</td>
<td>20</td>
<td>10^4</td>
</tr>
<tr>
<td>Absorbing Time Step</td>
<td>∆t_sim</td>
<td>µs</td>
<td>2</td>
<td>10</td>
</tr>
<tr>
<td># of Realizations</td>
<td>-</td>
<td>-</td>
<td>10^4</td>
<td>10^3</td>
</tr>
</tbody>
</table>

TABLE III
MAXIMUM VALUES USED TO NORMALIZE ALL CORRESPONDING CURVES.

<table>
<thead>
<tr>
<th>Signal</th>
<th>Symbol</th>
<th>Units</th>
<th>System 1</th>
<th>System 2</th>
</tr>
</thead>
<tbody>
<tr>
<td>Passive CIR</td>
<td>N^AB_RX</td>
<td>mol</td>
<td>24.7</td>
<td>48.4</td>
</tr>
<tr>
<td>Absorbing CIR</td>
<td>N^AB_RX</td>
<td>mol</td>
<td>1325</td>
<td>775</td>
</tr>
<tr>
<td>Passive ED</td>
<td>ED^AB</td>
<td>mol·s^-1</td>
<td>1.16 x 10^5</td>
<td>-</td>
</tr>
<tr>
<td>Absorption Rate</td>
<td>∆N^AB_RX</td>
<td>mol·s^-1</td>
<td>0.390</td>
<td>-</td>
</tr>
</tbody>
</table>

rate, we design the absorption rate detector \( \Delta N^A_B(\cdot) |^{AB} \) as
\[
\Delta N^A_B \left( t_m + \frac{\Delta t_M}{2} \right) |^{AB} = \frac{N^A_B(t_{m+1}) - N^A_B(t_m)}{\Delta t_M}, (27)
\]
which also applies to any environment, and we shift the observation time because (27) approximates the slope at the midpoint of the interval \([t_m, t_{m+1}]\).

V. SIMULATION AND NUMERICAL RESULTS

In this section, we verify the transform functions derived in Section III via simulations and numerical evaluation. We consider passive and absorbing receivers in two systems (one that is 3D and one that is 1D) that are summarized in Table II.

The simulations are completed using release v0.5 of the AcCoRD simulator [11] and averaged over the number of independent realizations listed in Table II. The realizations were sufficient to make confidence intervals negligible compared to deviations from the analytical expressions. The simulated 3D environment is truly unbounded, whereas the 1D environment is a 1 mm x 1 µm x 1 µm rectangular pipe that is centered at the RX. The simulations of the absorbing RX remove and count molecules if their trajectory in one simulation time step crosses the RX boundary. The sampling period \( \Delta t_M \) is double the time step \( \Delta t_{sim} \) used in passive RX simulations, so that there are samples at the times shifted by \( \frac{\Delta t_M}{2} \) for the implementation of the energy detector and absorption rate detector in (25) and (27), respectively. The time steps for simulating the absorbing RX are much smaller than for the passive RX so that we accurately model the absorbing reaction.

The results are presented in Figs. 1, 2, and 3. They all have the same format and are plotted on a log-log scale so that inaccuracies can be clearly observed. Figs. 1 and 3 display the passive CIR and the absorption rate (for Systems 1 and 2, respectively). Fig. 2 displays the passive energy detector and the absorbing CIR for System 1. All curves are grouped according to the signal that they correspond to, and are normalized by the maximum of the corresponding analytical curve (the values of which are listed in Table III).

For the specific equation used for each analytical curve, please refer to Table I. The “Simulation” curves are analogous to the “Analytical” curves, except that the receiver signals (including those used to calculate the energy and the absorption rate detector signals) are replaced by the corresponding average simulation output.

Fig. 1. Passive CIR and absorption rate of System 1 (3D) as a function of time. All curves are normalized to the maximum value of the corresponding analytical curve (solid black line), as listed in Table III. The transformed signals are found by applying (15) and (16).

Fig. 2. Passive energy detector (left) and absorbing CIR (right) of System 1 (3D). All curves are normalized to the maximum value of the corresponding analytical curve (solid black line), as listed in Table III. The transformed signals are found by applying (12) and (13).
All three figures show excellent agreement between the expected signals and those obtained via transforms, both analytically and using the simulations, so we will focus our discussion on a few key observations. First, we comment on the accuracy of the analytical transforms. All non-asymptotic transforms are very accurate for the entire range of time. They demonstrate that we have accurate means for transforming between the active and passive spherical receiver models. The accuracy of the asymptotic transforms greatly improves with time. They are not suitable for capturing the signal peak values, but may be appropriate for modeling intersymbol interference.

Second, we comment on the simulation accuracy. The accuracy of the absorbing RX signal depends on the simulation time step used. Even though we used time steps that were much smaller than those for the passive RX, the simulations underestimate the absorbing RX signal, particularly for small values of \( t \). This can be observed in the simulation of the absorbing signal in all three figures.

The simulated passive RX signals and the transforms that depend on those signals show deviations for very small values of \( t \) because we assumed that the concentration was uniform. However, these deviations become more accurate with increasing \( t \). In fact, the transforms of the passive signal become more accurate than the absorbing simulation. For example, for \( t = 0.003 \) s in Figs. 1 and 2, the 3D absorbing RX signal is more accurately obtained by transforming the passive simulation than by simulating the absorbing RX directly. If the same time step were used for both receiver models, then this improvement would be much more pronounced. Thus, for this system we can gain in both accuracy and computational efficiency by simulating a passive RX and transforming the results to apply to an active signal.

VI. CONCLUSIONS

In this paper, we studied the use of transforms to convert between the observed signal at a passive receiver and that at an absorbing receiver. We used two approaches to derive the transform functions for a spherical receiver in an unbounded diffusion-only system. We modified the weighted sum detector and the hitting rate detector to properly implement an integrator and a differentiator of the passive and absorbing impulse responses, respectively. Numerical results and simulations were shown to verify the transforms. Our future work will consider the derivation of transforms for more complex diffusion-based environments. We will also relax the use of the uniform concentration assumption to improve the accuracy in transforming the passive signal, and will consider the impact of the transforms on the channel statistics.

REFERENCES