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DOI:

[10.1109/TFUZZ.2018.2809686](https://doi.org/10.1109/TFUZZ.2018.2809686)

Document Version

Peer reviewed version

[Link to publication record in King's Research Portal](#)

Citation for published version (APA):

Wang, J. W., Tsai, S. H., Li, H. X., & Lam, H. K. (Accepted/In press). Spatially Piecewise Fuzzy Control Design for Sampled-Data Exponential Stabilization of Semi-linear Parabolic PDE Systems. *IEEE Transactions on Fuzzy Systems*. <https://doi.org/10.1109/TFUZZ.2018.2809686>

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Spatially Piecewise Fuzzy Control Design for Sampled-Data Exponential Stabilization of Semi-linear Parabolic PDE Systems

Jun-Wei Wang, Shun-Hung Tsai, Han-Xiong Li, and Hak-Keung Lam

Abstract—This paper employs a Takagi-Sugeno (T-S) fuzzy partial differential equation (PDE) model to solve the problem of sampled-data exponential stabilization in the sense of spatial L^∞ norm $\|\cdot\|_\infty$ for a class of nonlinear parabolic distributed parameter systems (DPSs), where only a few actuators and sensors are discretely distributed in space. Initially, a T-S fuzzy PDE model is assumed to be derived by the sector nonlinearity method to accurately describe complex spatiotemporal dynamics of the nonlinear DPSs. Subsequently, a static sampled-data fuzzy local state feedback controller is constructed based on the T-S fuzzy PDE model. By constructing an appropriate Lyapunov–Krasovskii functional candidate and employing vector-valued Wirtinger’s inequalities, a variation of vector-valued Poincaré–Wirtinger inequality in 1D spatial domain, as well as a vector-valued Agmon’s inequality, it is shown that the suggested sampled-data fuzzy controller exponentially stabilizes the nonlinear DPSs in the sense of $\|\cdot\|_\infty$, if sufficient conditions presented in term of standard linear matrix inequalities (LMIs) are fulfilled. Moreover, an LMI relaxation technique is utilized to enhance exponential stabilization ability of the suggested sampled-data fuzzy controller. Finally, the satisfactory and better performance of the suggested sampled-data fuzzy controller are demonstrated by numerical simulation results of two examples.

Index Terms—Sampled-data control, exponential stability, distributed parameter systems, Agmon’s inequality, Takagi-Sugeno fuzzy partial differential equation model.

I. INTRODUCTION

This work was supported in part by the National Natural Science Foundation of China under Grant 61403026, in part by the National Science Council of Taiwan, R.O.C., under Grant MOST-106-2221-E-027-009, in part by National Taipei University of Technology University of Science and Technology Beijing (NTUT-USTB) Joint Research Program under Grant NTUT-USTB-106-09 and the Fundamental Research Funds for the Central Universities under Grant FRF-TW201707, in part by the General Research Fund from Research Grants Council of Hong Kong SAR under Grant (CityU: 11205615) and the Hong Kong Scholars Program. (Corresponding authors: Jun-Wei Wang and Shun-Hung Tsai)

Jun-Wei Wang is with the Department of Systems Engineering and Engineering Management, City University of Hong Kong, Hong Kong SAR, P. R. China, and also with the School of Automation and Electrical Engineering, University of Science and Technology Beijing, Beijing 100083, P. R. China, e-mail: junweiwang@ustb.edu.cn (J.-W. Wang)

Shun-Hung Tsai is with the Graduate Institute of Automation Technology, National Taipei University of Technology, Taipei, Taiwan, R.O.C., e-mail: shstai@ntut.edu.tw (S.-H. Tsai)

Han-Xiong Li is with the Department of Systems Engineering and Engineering Management, City University of Hong Kong, Hong Kong SAR, P. R. China, and also with the State Key Laboratory of High Performance Complex Manufacturing, Central South University, Changsha, Hunan, P. R. China, e-mail: mehqli@cityu.edu.hk (H.-X. Li)

Hak-Keung Lam is with the Department of Informatics, King’s College London, London, U.K., e-mail: hak-keung.lam@kcl.ac.uk (H.-K. Lam).

MUCH research has been done concerned with sampled-data control systems over the past few decades, leading to considerable results for analysis and synthesis of sampled-data control systems (e.g., [1]–[13] and the references therein). Note that the aforementioned results are mainly developed for dynamic systems whose behavior only depends on time and mathematical models take the form of ordinary/delay differential equations (O/DDEs). In real industrial applications, however, most processes are spatiotemporal in nature so that their behavior must depend on time as well as spatial position, for example, thermal diffusion processes, pipe flow, chemical processes, and flexible mechanical structure, etc [14]–[17]. These spatiotemporal processes are modeled as distributed parameter systems (DPSs) described by partial differential equations (PDEs). By the model reduction techniques (e.g., Galerkin’s method), a finite-dimensional sampled-data boundary control design was developed in [18] for a linear parabolic DPS and a finite-dimensional observer-based active fault-tolerant control design was reported in [19] for nonlinear DPSs with sampled-data measurements in time. But an exact performance of closed-loop DPSs is difficultly achieved due to truncation before control design. To achieve exact and better control performance, some infinite-dimensional robust design results have been developed in [20]–[23] for linear sampled-data controllers of nonlinear parabolic DPSs directly on the basis of their original semi-linear PDE model.

On the other hand, fuzzy control based on the Takagi-Sugeno (T-S) fuzzy model [24] offers a conceptually simple, systematic and effective framework to support nonlinear control design of complex dynamic systems and has become increasingly popular during the past few decades (see, e.g., [25]–[27]). In the last decade, many fruitful results of fuzzy-model-based sampled-data control design have been reported for nonlinear O/DDE systems [28]–[38]. With the help of the Galerkin’s method with the singular perturbation approach, finite-dimensional guaranteed cost and H_∞ fuzzy-model-based sampled-data control designs have been recently proposed in [39] and [40] for semi-linear parabolic PDE (SLP-PDE) systems. Although fuzzy-model-based control approach has been successfully extended in [41]–[46] for infinite-dimensional nonlinear control design of SLPPDE systems by proposing T-S fuzzy PDE model, the suggested results mainly focus on exponential stabilization of SLPPDE systems via continuous-time fuzzy controllers. To the best of authors’ knowledge, the result on infinite-dimensional fuzzy-model-based control design has not been reported yet for sampled-

data exponential stabilization of SLPPDE systems with a finite number of actuators and sensors, which motivates this study.

This paper discusses the problem of infinite-dimensional fuzzy-model-based sampled-data control design in the sense of spatial L^∞ norm $\|\cdot\|_\infty$ for a class of SLPPDE systems with a finite number of actuators and sensors discretely distributed over the spatial domain. The objective of this paper is to develop a conceptually simple but effective infinite-dimensional design method of a sampled-data fuzzy controller only using state information taken from sensors located at some known local areas of the spatial domain (i.e., local piecewise state information), such that the resulting closed-loop PDE system is exponentially stable in the sense of $\|\cdot\|_\infty$. In the proposed design method, a T-S fuzzy PDE model is first assumed to be constructed [45], [46] to accurately describe the complex spatiotemporal dynamics of the SLPPDE system. Based on the fuzzy PDE model, a sampled-data fuzzy controller is then constructed only using local piecewise state information. It is shown by constructing an appropriate Lyapunov-Krasovskii functional candidate and using vector-valued Wirtinger's inequalities, a variation of vector-valued Poincaré-Wirtinger inequality in 1D spatial domain, as well as a vector-valued Agmon's inequality that the suggested sampled-data fuzzy controller exponentially stabilizes the SLPPDE system in the sense of $\|\cdot\|_\infty$. Moreover, exponential stabilization ability of the suggested sampled-data fuzzy controller is enhanced by providing an LMI relaxation technique. The main results of this paper are presented in terms of standard LMIs, which are directly solved by the solvers in MATLAB's LMI Control Toolbox [47]. Finally, the effectiveness and merit of the suggested sampled-data fuzzy controller are demonstrated by numerical simulation results of two examples.

Main contribution and novelty of this paper are summarized as follows: (i) A variation of vector-valued Poincaré-Wirtinger inequality in 1D spatial domain is provided by first mean value theorem for definite integrals; (ii) A vector-valued Agmon's inequality is introduced from the scalar one; (iii) An LMI-based infinite-dimensional design method of static sampled-data fuzzy controller is developed for exponential stabilization of the SLPPDE system in the sense of $\|\cdot\|_\infty$. In comparison to existing finite-dimensional fuzzy sampled-data control designs in [39] and [40] for the SLPPDE systems, the main difference of the proposed design method of this paper lies in that finite-dimensional fuzzy sampled-data control designs in [39] and [40] only ensure exponential stabilization in the sense of spatial L^2 norm $\|\cdot\|_2$, while the infinite-dimensional sampled-data fuzzy control design of this paper guarantees exponential stabilization in the sense of $\|\cdot\|_\infty$.

The organization of the rest of this paper is given as follows. Section II provides preliminaries and problem formulation. Section III gives a static sampled-data fuzzy control design. Numerical simulation results are provided in Section IV to show satisfactory and better performance of the suggested sampled-data fuzzy controller than the existing ones. Finally, Section V offers some brief concluding remarks.

Notation: The set of all real numbers, n -dimensional Euclidean space with the norm $\|\cdot\|$ and the set of all $m \times n$ matrices are denoted by $\mathbb{R}$, $\mathbb{R}^n$ and $\mathbb{R}^{m \times n}$, respectively. $|\varsigma|$

denotes absolute value of any real number ς . $\mathbf{I}$ stands for an identity matrix of appropriate dimension. For a matrix $\mathbf{M} \in \mathbb{R}^{n \times n}$, $\mathbf{M} > (<, \leq) 0$ means that it is symmetric and positive definite (negative definite, negative semi-definite, respectively). For a square matrix $\mathbf{A}$, its minimum and maximum eigenvalues are denoted by $\lambda_{\min}(\mathbf{A})$ and $\lambda_{\max}(\mathbf{A})$, respectively. For a given scalar $L > 0$, $\mathcal{L}_n^2([0, L]) \triangleq \mathcal{L}^2([0, L]; \mathbb{R}^n)$ is a Hilbert space of square integrable vector functions $\omega(x) : [0, L] \rightarrow \mathbb{R}^n$ with the norm $\|\omega(\cdot)\|_2 \triangleq \sqrt{\int_0^L \omega^T(x) \omega(x) dx}$. This norm is also referred to as *spatial L^2 norm*. Given an integer $\bar{k}$ and a constant $L > 0$, $\mathcal{H}_n^{\bar{k}}((0, L)) \triangleq \mathcal{W}^{\bar{k}, 2}((0, L); \mathbb{R}^n)$ is a Sobolev space of absolutely continuous vector functions $\omega(x) : (0, L) \rightarrow \mathbb{R}^n$ with square integrable derivatives $\frac{d^{\bar{k}} \omega(x)}{dx^{\bar{k}}}$ of the order $\bar{k}$ and with the norm $\|\omega(\cdot)\|_{\mathcal{H}_n^{\bar{k}}((0, L))} \triangleq \sqrt{\int_0^L \sum_{i=0}^{\bar{k}} \frac{d^i \omega^T(x)}{dx^i} \frac{d^i \omega(x)}{dx^i} dx}$. For any $\mathbf{y}(\cdot, t) \in \mathcal{H}_n^1((0, L))$, *spatial L^∞ norm* is defined as $\|\mathbf{y}(\cdot, t)\|_\infty \triangleq \max_{x \in [0, L]} \|\mathbf{y}(x, t)\|$. The subscripts x and t of $\mathbf{y}(x, t)$ are partial derivatives with respect to x and t , i.e., $\mathbf{y}_t(x, t) = \partial \mathbf{y}(x, t) / \partial t$, $\mathbf{y}_x(x, t) = \partial \mathbf{y}(x, t) / \partial x$, and $\mathbf{y}_{xx}(x, t) = \partial^2 \mathbf{y}(x, t) / \partial x^2$, respectively. The transpose of a vector or a matrix is denoted by the superscript ' T '. The symbol '*' in matrix expressions denotes an ellipsis induced by symmetry.

II. PRELIMINARIES AND PROBLEM FORMULATION

A. System description

The class of nonlinear DPSs under consideration is described by the following SLPPDE in 1D spatial domain

$$\begin{cases} \mathbf{y}_t(x, t) = \Theta \mathbf{y}_{xx}(x, t) + \mathbf{f}(\mathbf{y}(x, t)) \\ \quad + \mathbf{G}(x) \mathbf{u}(t), \quad x \in (0, L), \quad t > 0, \\ \mathbf{y}(0, t) = 0, \quad \mathbf{y}_x(x, t)|_{x=L} = 0, \quad t > 0, \\ \mathbf{y}(x, 0) = \mathbf{y}_0(x), \quad x \in [0, L], \end{cases} \quad (1)$$

where $\mathbf{y}(x, t) \triangleq [y_1(x, t) \ y_2(x, t) \ \cdots \ y_n(x, t)]^T \in \mathbb{D}$ is state ($\mathbb{D} \triangleq \{\mathbf{y}(\cdot, t) \in \mathcal{L}_n^2([0, L]) \mid \|\mathbf{y}_i(\cdot, t)\|_\infty \leq \phi_i, \ i \in \{1, 2, \dots, n\}\} \subset \mathcal{L}_n^2([0, L])$ is a given local domain containing the equilibrium profile $\mathbf{y}(\cdot, t) = 0$, $\phi_i > 0$, $i \in \{1, 2, \dots, n\}$ are real known scalars), $x \in [0, L] \subset \mathbb{R}$ and $t \geq 0$ are spatial position and time variables, respectively. $0 < \Theta \in \mathbb{R}^{n \times n}$ is a known diffusivity coefficient matrix and controls how fast the masses can spread in the media, the term $\Theta \mathbf{y}_{xx}(x, t)$ describes diffusion phenomenon and that is a spontaneous dispersion of mass in the spatial domain from the higher density/concentration/temperature area to the lower density/concentration/temperature area, $\mathbf{f}(\mathbf{y}(x, t))$ is sufficiently smooth in $\mathbf{y}(x, t)$ and satisfies $\mathbf{f}(0) = 0$. $\mathbf{u}(t) \triangleq [u_1(t) \ u_2(t) \ \cdots \ u_m(t)]^T \in \mathbb{R}^m$ is control input provided by m actuators, which are discretely distributed over the spatial domain $(0, L)$. $\mathbf{G}(x)$ is a known square integrable matrix function of x , in which the i -th column describes the i -th actuator's distribution. Here, $\mathbf{G}(x)$ is chosen as $\mathbf{G}(x) \triangleq [\mathbf{g}_1(x) \ \mathbf{g}_2(x) \ \cdots \ \mathbf{g}_m(x)] \in \mathbb{R}^{n \times m}$, where

$$\mathbf{g}_v(x) \triangleq \begin{cases} \mathbf{g}_v & x \in [x_v, x_{v+1}) \\ 0 & \text{otherwise,} \end{cases} \quad v \in \mathcal{M}, \quad (2)$$

Fig. 1: Spatial piecewise control form [46]

with $[x_v, x_{v+1})$ is the v -th actuator coverage area and $\mathcal{M} \triangleq \{1, 2, \dots, m\}$. This form $\mathbf{G}(x)$ will produce m zones of *spatial piecewise control* over the intervals $[x_v, x_{v+1})$, $v \in \mathcal{M}$ [46], which is illustrated in Fig. 1. It is easily seen from Fig. 1 that $0 = x_1 < x_2 < \dots < x_m < x_{m+1} = L$.

We introduce the following definitions of exponential stability in the sense of $\|\cdot\|_2$ and $\|\cdot\|_\infty$:

Definition 1: (Exponential stability in the sense of $\|\cdot\|_2$, [46]) The SLPPDE system (1) with $\mathbf{u}(t) \equiv 0$ is said to be *exponentially stable in the sense of $\|\cdot\|_2$* , if there exist two constants $\sigma \geq 1$ and $\rho > 0$ such that $\|\mathbf{y}(\cdot, t)\|_2 \leq \sigma \|\mathbf{y}_0(\cdot)\|_2 \exp(-\rho t)$ is fulfilled for any $t \geq 0$.

Definition 2: (Exponential stability in the sense of $\|\cdot\|_\infty$) The SLPPDE system (1) with $\mathbf{u}(t) \equiv 0$ is said to be *exponentially stable in the sense of $\|\cdot\|_\infty$* , if there exist two constants $\alpha \geq 1$ and $\beta > 0$ such that $\|\mathbf{y}(\cdot, t)\|_\infty \leq \alpha \exp(-\beta t)$ is fulfilled for any $t \geq 0$.

Remark 1: Clearly, Definitions 1 and 2 are not equivalent and Definition 2 is stronger than Definition 1 as if the system (1) is exponentially stable in the sense of $\|\cdot\|_\infty$, it must be exponentially stable in the sense of $\|\cdot\|_2$, but not vice versa.

B. T-S fuzzy PDE model and problem formulation

For the convenience of fuzzy control design, by following the main idea of the sector nonlinearity method for semi-linear PDE systems [45] and [46], we assume that the following T-S fuzzy PDE model is constructed by the sector nonlinearity method to exactly describe complex spatiotemporal dynamics for the SLPPDE of (1) in a given operating domain $\mathfrak{S}$:

Plant Rule i :

IF $\xi_1(x, t)$ is F_{i1} and $\dots$ and $\xi_l(x, t)$ is F_{il} , **THEN**

$$\begin{aligned} \mathbf{y}_t(x, t) = & \Theta \mathbf{y}_{xx}(x, t) + \mathbf{A}_i \mathbf{y}(x, t) \\ & + \mathbf{G}(x) \mathbf{u}(t), \quad x \in (0, L), \quad t > 0, \quad i \in \mathbb{S} \end{aligned} \quad (3)$$

where F_{ij} , $i \in \mathbb{S} \triangleq \{1, 2, \dots, r\}$, $j \in \{1, 2, \dots, l\}$ are fuzzy sets, $\mathbf{A}_i \in \mathbb{R}^{n \times n}$, $i \in \mathbb{S}$ are known matrices, and $r = 2^l$ is the number of IF-THEN fuzzy rules. The premise variables $\xi_1(x, t)$, $\xi_2(x, t)$, $\dots$, $\xi_l(x, t)$ are assumed to be functions of $\mathbf{y}(x, t)$. The overall dynamics of the fuzzy parabolic PDE (3) is expressed as

$$\begin{aligned} \mathbf{y}_t(x, t) = & \Theta \mathbf{y}_{xx}(x, t) + \sum_{i=1}^r h_i(\boldsymbol{\xi}(x, t)) \mathbf{A}_i \mathbf{y}(x, t) \\ & + \mathbf{G}(x) \mathbf{u}(t), \quad x \in (0, L), \quad t > 0, \end{aligned} \quad (4)$$

where $\boldsymbol{\xi}(x, t) \triangleq [\xi_1(x, t) \quad \xi_2(x, t) \quad \dots \quad \xi_l(x, t)]^T$ and $h_i(\boldsymbol{\xi}(x, t)) \triangleq \frac{\prod_{j=1}^l F_{ij}(\xi_j(x, t))}{\sum_{i=1}^r \prod_{j=1}^l F_{ij}(\xi_j(x, t))}$, $i \in \mathbb{S}$, $F_{ij}(\xi_j(x, t))$ is

grade of the membership of $\xi_j(x, t)$ in F_{ij} for $i \in \mathbb{S}$. In this paper, it is assumed that $\prod_{j=1}^l F_{ij}(\xi_j(x, t)) \geq 0$, $i \in \mathbb{S}$ and $\sum_{i=1}^r \prod_{j=1}^l F_{ij}(\xi_j(x, t)) > 0$ for $x \in (0, L)$ and $t > 0$. Hence, $h_i(\boldsymbol{\xi}(x, t))$, $i \in \mathbb{S}$ have the following property for $x \in (0, L)$ and $t > 0$:

$$h_i(\boldsymbol{\xi}(x, t)) \geq 0, \quad i \in \mathbb{S} \quad \text{and} \quad \sum_{i=1}^r h_i(\boldsymbol{\xi}(x, t)) = 1. \quad (5)$$

From above analysis, we know that the SLPPDE in (1) is equivalent to the fuzzy PDE (4) in the operating domain $\mathfrak{S}$. Generally, the operating domain $\mathfrak{S}$ is chosen as $\mathbb{D} \subset \mathfrak{S}$ to guarantee the *robustness* of the fuzzy PDE model (4).

This study considers a static sampled-data fuzzy local state feedback controller of the following form by the fuzzy PDE model (4):

$$\begin{aligned} u_v(t) = & \sum_{i=1}^r h_{vi}(t_k) \frac{\mathbf{k}_{vi}^T}{\Delta \tilde{x}_v} \int_{\tilde{x}_{2v-1}}^{\tilde{x}_{2v}} \mathbf{y}(x, t_k) dx, \quad v \in \mathcal{M}, \\ & t \in [t_k, t_{k+1}), \quad k \in \wp \triangleq \{0, 1, 2, \dots\}, \end{aligned} \quad (6)$$

where

$$\begin{aligned} h_{vi}(t_k) \triangleq & \frac{1}{\Delta \tilde{x}_v} \int_{\tilde{x}_{2v-1}}^{\tilde{x}_{2v}} h_i(\boldsymbol{\xi}(x, t_k)) dx, \\ \Delta \tilde{x}_v \triangleq & \tilde{x}_{2v} - \tilde{x}_{2v-1}, \quad v \in \mathcal{M}, \end{aligned} \quad (7)$$

and $\mathbf{k}_{vi} \in \mathbb{R}^n$, $v \in \mathcal{M}$, $i \in \mathbb{S}$ are control gain parameters to be determined, and t_k , $k \in \wp$ are sampling instants and satisfy $t_{k+1} - t_k \leq T_h$, $k \in \wp$ ($T_h > 0$ is the maximum sampling step in time given in advance). From (5) and (7), we get

$$h_{vi}(t_k) \geq 0, \quad i \in \mathbb{S} \quad \text{and} \quad \sum_{i=1}^r h_{vi}(t_k) = 1, \quad (8)$$

for any sampling instants t_k , $k \in \wp$ and $v \in \mathcal{M}$. Clearly, the sampled-data fuzzy control law (6) with (7) is a weighted average of sampled-data linear control laws $\frac{\mathbf{k}_{vi}^T}{\Delta \tilde{x}_v} \int_{\tilde{x}_{2v-1}}^{\tilde{x}_{2v}} \mathbf{y}(x, t_k) dx$, $v \in \mathcal{M}$, $i \in \mathbb{S}$, $k \in \wp$, where the mean value of the sampled-data state $\mathbf{y}(x, t_k)$, $k \in \wp$ is utilized and the weights are determined by the functions $h_{vi}(t_k)$, $v \in \mathcal{M}$, $i \in \mathbb{S}$, $k \in \wp$ (i.e., the mean value of fuzzy membership functions $h_i(\boldsymbol{\xi}(x, t))$, $i \in \mathbb{S}$ on the subintervals $[\tilde{x}_{2v-1}, \tilde{x}_{2v}]$, $v \in \mathcal{M}$ at the sampling instants t_k , $k \in \wp$).

By substituting the sampled-data fuzzy controller (6) into the fuzzy PDE (4), considering the boundary conditions and the initial condition in (1), and using the definition of $\mathbf{G}(x)$ given in (2), the resulting closed-loop sampled-data fuzzy PDE system is given as follows:

$$\begin{cases} \mathbf{y}_t(x, t) = \Theta \mathbf{y}_{xx}(x, t) + \sum_{i=1}^r h_i(\boldsymbol{\xi}(x, t)) \mathbf{A}_i \mathbf{y}(x, t) \\ \quad + \mathbf{g}_v \sum_{i=1}^r h_{vi}(t_k) \frac{\mathbf{k}_{vi}^T}{\Delta \tilde{x}_v} \int_{\tilde{x}_{2v-1}}^{\tilde{x}_{2v}} \mathbf{y}(x, t_k) dx, \\ \quad k \in \wp, \quad x \in [x_v, x_{v+1}), \quad t > 0, \quad v \in \mathcal{M}, \\ \mathbf{y}(0, t) = 0, \quad \mathbf{y}_x(x, t)|_{x=L} = 0, \quad t > 0, \\ \mathbf{y}(x, 0) = \mathbf{y}_0(x), \quad x \in [0, L]. \end{cases} \quad (9)$$

The fuzzy-model-based sampled-data control design addressed in this paper is formally defined as follows:

For the SLPPDE system (1), the objective of this study is to develop an LMI-based design of the static sampled-data fuzzy controller (6) such that the resulting closed-loop SLPPDE system is exponentially stable in the sense of $\|\cdot\|_\infty$.

C. Important lemmas

Lemma 1 (Vector-valued Wirtinger's inequalities): Let $\mathbf{y}(\cdot, t) \in \mathcal{H}_n^1((0, L))$ be a vector function with $\mathbf{y}(0, t) = 0$ or $\mathbf{y}(L, t) = 0$, $t \geq 0$. Then, for any matrix $0 \leq \mathbf{S} \in \mathbb{R}^{n \times n}$, we have for any $t \geq 0$

$$\int_0^L \mathbf{y}^T(x, t) \mathbf{S} \mathbf{y}(x, t) dx \leq \frac{4L^2}{\pi^2} \int_0^L \mathbf{y}_x^T(x, t) \mathbf{S} \mathbf{y}_x(x, t) dx. \quad (10)$$

Moreover, for a vector function $\mathbf{y}(\cdot, t) \in \mathcal{H}_n^2((0, L))$ with $\mathbf{y}_x(0, t) = 0$ or $\mathbf{y}_x(L, t) = 0$, $t \geq 0$, we have the following inequality for any matrix $0 \leq \mathbf{S} \in \mathbb{R}^{n \times n}$

$$\begin{aligned} \int_0^L \mathbf{y}_x^T(x, t) \mathbf{S} \mathbf{y}_x(x, t) dx \\ \leq \frac{4L^2}{\pi^2} \int_0^L \mathbf{y}_{xx}^T(x, t) \mathbf{S} \mathbf{y}_{xx}(x, t) dx, \quad t \geq 0. \end{aligned} \quad (11)$$

Proof: The proof of this lemma is easily done by Theorem 2 in [48] and thus omitted here. ■

Lemma 2 (A variation of vector-valued Poincaré–Wirtinger inequality in 1D spatial domain): Let $\mathbf{y}(\cdot, t) \in \mathcal{H}_n^1((0, L))$, $t \geq 0$ be a vector function. Then, for any matrix $0 \leq \mathbf{S} \in \mathbb{R}^{n \times n}$, we have

$$\begin{aligned} \int_0^L (\mathbf{y}(x, t) - \mathbf{y}_l(t))^T \mathbf{S} (\mathbf{y}(x, t) - \mathbf{y}_l(t)) dx \\ \leq 4\phi\pi^{-2} \int_0^L \mathbf{y}_x^T(x, t) \mathbf{S} \mathbf{y}_x(x, t) dx, \quad t \geq 0 \end{aligned} \quad (12)$$

where $\mathbf{y}_l(t) \triangleq (l_2 - l_1)^{-1} \int_{l_1}^{l_2} \mathbf{y}(x, t) dx$, $[l_1, l_2] \subset [0, L]$, and $\phi \triangleq \max\{l_2^2, (L - l_1)^2\}$.

Proof: See the Appendix A ■

Lemma 3 (Vector-valued Agmon's inequality): Let $\mathbf{y}(\cdot, t) \in \mathcal{H}_n^1((0, L))$ be a vector function with $\mathbf{y}(0, t) = 0$ or $\mathbf{y}(L, t) = 0$, $t \geq 0$. The following inequality is fulfilled:

$$\|\mathbf{y}(\cdot, t)\|_\infty^2 \leq 2\|\mathbf{y}(\cdot, t)\|_2 \|\mathbf{y}_x(\cdot, t)\|_2 \leq \|\mathbf{y}(\cdot, t)\|_{\mathcal{H}_n^1((0, L))}^2, \quad t \geq 0. \quad (13)$$

Proof: See the Appendix B ■

Note that the membership functions $h_{vi}(t_k)$, $v \in \mathcal{M}$, $i \in \mathbb{S}$, $k \in \wp$ in the sampled-data fuzzy controller (6) are different from the ones $h_i(\xi(x, t))$, $i \in \mathbb{S}$ in the fuzzy PDE model (4). This characteristic prevents direct use of the parameterized LMI technique [41] to reduce control design conservativeness. To overcome this drawback, this paper will provide an LMI relaxation technique. By considering (5) and (8), we get

$$|h_i(\xi(x, t)) - h_{vi}(t_k)| \leq 1, \quad x \in [x_v, x_{v+1}), \quad v \in \mathcal{M} \quad (14)$$

for any $t \in [t_k, t_{k+1})$, $k \in \wp$, and $i \in \mathbb{S}$. Using (14), the following lemma provides an LMI relaxation technique:

Lemma 4: Consider $n \times n$ matrices $\mathbf{\Lambda}_v(x, t, t_k) \triangleq \sum_{i=1}^r \sum_{j=1}^r h_i(\xi(x, t)) h_{vj}(t_k) \mathbf{\Lambda}_{vij}$, $x \in [x_v, x_{v+1})$, $v \in \mathcal{M}$, $t \in [t_k, t_{k+1})$, $k \in \wp$. The matrix inequalities $\mathbf{\Lambda}_v(x, t, t_k) < 0$, $v \in \mathcal{M}$ are fulfilled for $x \in [x_v, x_{v+1})$, $v \in \mathcal{M}$, $t \in [t_k, t_{k+1})$, $k \in \wp$, if there exist matrices $0 > \mathbf{\Gamma}_{vij} \in \mathbb{R}^{n \times n}$, $0 > \mathbf{\Upsilon}_{vij} \in$

$\mathbb{R}^{n \times n}$, $\mathbf{\Omega}_{vij} = \mathbf{\Omega}_{vji}^T \in \mathbb{R}^{n \times n}$, and $\mathbf{\Omega}_{vi(j+r)} = \mathbf{\Omega}_{v(j+r)i}^T \in \mathbb{R}^{n \times n}$, $v \in \mathcal{M}$, $i, j \in \mathbb{S}$ such that

$$\mathbf{\Delta}_{vij} + \mathbf{\Delta}_{vji} \leq \mathbf{\Omega}_{vij} + \mathbf{\Omega}_{vji}, \quad v \in \mathcal{M}, \quad i, j \in \mathbb{S} \quad (15)$$

$$\begin{aligned} \mathbf{\Lambda}_{vij} - 2\mathbf{\Delta}_{vij} - \sum_{k=1}^r (\mathbf{\Delta}_{vik}^+ + \mathbf{\Delta}_{vki}^+) \\ \leq \mathbf{\Omega}_{vi(j+r)} + \mathbf{\Omega}_{v(j+r)i}, \quad v \in \mathcal{M}, \quad i, j \in \mathbb{S} \end{aligned} \quad (16)$$

$$\mathbf{Y}_v \triangleq \begin{bmatrix} \mathbf{Y}_{v11} & \mathbf{Y}_{v12} \\ * & \mathbf{Y}_{v11} \end{bmatrix}, \quad v \in \mathcal{M} \quad (17)$$

where $\mathbf{\Delta}_{vij} \triangleq \mathbf{\Gamma}_{vij} - \mathbf{\Upsilon}_{vij}$, $\mathbf{\Delta}_{v,i,j}^+ \triangleq \mathbf{\Gamma}_{vij} + \mathbf{\Upsilon}_{vij}$, $i, j \in \mathbb{S}$,

$$\mathbf{Y}_{v,11} \triangleq \begin{bmatrix} \mathbf{\Omega}_{v11} & \cdots & \mathbf{\Omega}_{v1r} \\ \vdots & \ddots & \vdots \\ \mathbf{\Omega}_{vr1} & \cdots & \mathbf{\Omega}_{vrr} \end{bmatrix}, \quad v \in \mathcal{M},$$

$$\mathbf{Y}_{v,12} \triangleq \begin{bmatrix} \mathbf{\Omega}_{v1(r+1)} & \cdots & \mathbf{\Omega}_{v1(2r)} \\ \vdots & \ddots & \vdots \\ \mathbf{\Omega}_{vr(r+1)} & \cdots & \mathbf{\Omega}_{vr(2r)} \end{bmatrix}, \quad v \in \mathcal{M}.$$

Proof: The proof of this lemma is easily done via Lemma 2 in [46] and thus omitted here. ■

Remark 2: LMI relaxation techniques have been proposed in [45] and [46] by making assumptions on fuzzy membership functions to reduce the fuzzy control design conservativeness. These assumptions to some extent restrict the application of the control design in [45] and [46]. In this paper, an LMI relaxation technique is developed in Lemma 4 by removing these assumptions made in [45] and [46]. Lemma 4 is a special case of the LMI relaxation technique in [46], where the positive parameters $\varsigma_{v,i}$ are chosen as 1.

III. STATIC SAMPLED-DATA FUZZY CONTROL DESIGN

Let us consider a Lyapunov–Krasovskii functional candidate of the following form for the system (9):

$$V(t) = V_1(t) + V_2(t) + V_3(t), \quad t \in [t_k, t_{k+1}), \quad k \in \wp, \quad (18)$$

where

$$V_1(t) = \int_0^L \mathbf{y}^T(x, t) \mathbf{P}_1 \mathbf{y}(x, t) dx, \quad (19)$$

$$V_2(t) = \int_0^L \mathbf{y}_x^T(x, t) \mathbf{P}_2 \mathbf{y}_x(x, t) dx, \quad (20)$$

$$V_3(t) = T_h \int_0^L \int_{t_k}^t (s - t + T_h) \mathbf{y}_s^T(x, s) \mathbf{P}_3 \mathbf{y}_s(x, s) ds dx, \quad (21)$$

with $0 < \mathbf{P}_1 \in \mathbb{R}^{n \times n}$, $0 < \mathbf{P}_2 \in \mathbb{R}^{n \times n}$, and $0 < \mathbf{P}_3 \in \mathbb{R}^{n \times n}$ are Lyapunov matrices to be determined. Obviously, the function $V(t)$ given in (18) is continuous in time from the right for $\mathbf{y}(x, t)$ satisfying the system (9) and absolutely continuous for $t \neq t_k$, $k \in \wp$. Along the jumps t_k , $k \in \wp$ $V(t)$ satisfies $V(t_k) \leq V(t_k^-) \triangleq \lim_{t \rightarrow t_k^-} V(t)$, $k \in \wp$ since $V_3(t_k^-) \triangleq \lim_{t \rightarrow t_k^-} V_3(t) \geq 0$ and $V_3(t_k) = 0$ after the jumps

because $\mathbf{y}(x, t)|_{t=t_k} = \mathbf{y}(x, t_k)$. Moreover, $V(t)$ satisfies the following inequality for any $t \in [t_k, t_{k+1})$, $k \in \mathcal{S}$:

$$\begin{aligned} \rho_1 \|\mathbf{y}(\cdot, t)\|_{\mathcal{H}_n^1((0, L))}^2 &\leq V(t) \leq \lambda_{\max}(\mathbf{P}_1) \int_0^L \mathbf{y}^T(x, t) \mathbf{y}(x, t) dx \\ &+ \lambda_{\max}(\mathbf{P}_2) \int_0^L \mathbf{y}_x^T(x, t) \mathbf{y}_x(x, t) dx \\ &+ \lambda_{\max}(\mathbf{P}_3) T_h^2 \\ &\times \int_0^L \int_{t_k}^t \mathbf{y}_s^T(x, s) \mathbf{y}_s(x, s) ds dx, \end{aligned} \quad (22)$$

where $\rho_1 \triangleq \min\{\lambda_{\min}(\mathbf{P}_1), \lambda_{\min}(\mathbf{P}_2)\}$. For $t = 0$, the following inequality can be obtained from (22)

$$V(0) \leq \rho_2 \|\mathbf{y}_0(\cdot)\|_{\mathcal{H}_n^1((0, L))}^2 \quad (23)$$

where $\rho_2 \triangleq \max\{\lambda_{\max}(\mathbf{P}_1), \lambda_{\max}(\mathbf{P}_2)\}$.

Note that the proposed Lyapunov-Krasovskii functional candidate of the form (18)-(21) is different from that reported in [3]- [6], [9]- [11], and [28]- [38] for the sampled-data control design of the systems which are modeled by O/DDEs, due to that the system (1) is spatiotemporal and includes the term $\Theta \mathbf{y}_{xx}(x, t)$. Although the problem of sampled-data fuzzy control design of nonlinear parabolic DPSs has been addressed in [39] and [40], the proposed Lyapunov-Krasovskii functional candidate is constructed based on the low-dimensional ODE approximations obtained from the singular perturbation formulation of Galerkin's method. On the other hand, the Lyapunov-Krasovskii functional candidate of the form (18)-(21) is also different from the ones used in [20]- [22] for sampled-data control design of semi-linear parabolic PDE systems, where an exponential term $\exp(2\alpha(s - t))$ (α is a given positive constant) is introduced in construction of Lyapunov-Krasovskii functional candidates.

Theorem 1: Consider the SLPPDE system (1) and the T-S fuzzy PDE model (4). For given constants $L > 0$, $T_h > 0$, an integer $m > 0$, and parameters $\tilde{x}_{2v-1}$, $\tilde{x}_{2v}$, $v \in \mathcal{M}$, $x_2, x_3, \dots, x_m$, if there exist matrices $0 < \mathbf{X}_1 \in \mathbb{R}^{n \times n}$, $0 < \mathbf{X}_2 \in \mathbb{R}^{n \times n}$, $0 < \mathbf{X}_3 \in \mathbb{R}^{n \times n}$ and vectors $\mathbf{z}_{vj} \in \mathbb{R}^n$, $v \in \mathcal{M}$, $j \in \mathbb{S}$ such that the following LMIs are fulfilled:

$$\Phi_{vij} \triangleq \begin{bmatrix} \Phi_{1vij} & \Phi_{2vij} \\ * & \Phi_{3vij} \end{bmatrix} < 0, \quad v \in \mathcal{M}, \quad i, j \in \mathbb{S} \quad (24)$$

where

$$\begin{aligned} \Phi_{1vij} &\triangleq \begin{bmatrix} [\mathbf{A}_i \mathbf{X}_1 + *] - \frac{\pi^2}{4\varphi_v} \Phi_1 & \frac{\pi^2}{4\varphi_v} \Phi_1 + \mathbf{g}_v \mathbf{z}_{vj}^T \\ * & -\frac{\pi^2}{4\varphi_v} \Phi_1 \end{bmatrix}, \\ \Phi_{2vij} &\triangleq \begin{bmatrix} -\mathbf{X}_1 \mathbf{A}_i^T & \mathbf{X}_1 \mathbf{A}_i^T & -\mathbf{g}_v \mathbf{z}_{vj}^T \\ -\mathbf{z}_{vj} \mathbf{g}_v^T & \mathbf{z}_{vj} \mathbf{g}_v^T & 0 \end{bmatrix}, \\ \Phi_{3vij} &\triangleq \begin{bmatrix} -[\Theta \mathbf{X}_2 + *] & \mathbf{X}_2 \Theta & \mathbf{g}_v \mathbf{z}_{vj}^T \\ * & T_h^2 \mathbf{X}_3 - 2\mathbf{X}_1 & -\mathbf{g}_v \mathbf{z}_{vj}^T \\ * & * & -\frac{\Delta \tilde{x}_v}{\Delta x_v} \mathbf{X}_3 \end{bmatrix}, \end{aligned}$$

in which

$$\begin{aligned} \Phi_1 &\triangleq [\Theta \mathbf{X}_1 + *], \\ \varphi_v &\triangleq \max\{(\tilde{x}_{2v} - x_v)^2, (x_{v+1} - \tilde{x}_{2v-1})^2\}, \\ \Delta x_v &\triangleq x_{v+1} - x_v, \quad v \in \mathcal{M}, \end{aligned} \quad (25)$$

then there exists a static sampled-data fuzzy controller (6) such that the resulting closed-loop system is exponentially stable in the sense of $\|\cdot\|_\infty$, where the control gain parameters $\mathbf{k}_{vj}$, $v \in \mathcal{M}$, $j \in \mathbb{S}$ are given by

$$\mathbf{k}_{vj}^T = \mathbf{z}_{vj}^T \mathbf{X}_1^{-1}, \quad v \in \mathcal{M}, \quad j \in \mathbb{S}. \quad (26)$$

Proof: See the Appendix C. ■

By the Lyapunov-Krasovskii functional candidate (18) with (19)-(21) and making use of integration by parts and Lemmas 1-3, Theorem 1 provides an LMI-based design of static sampled-data fuzzy controller (6) exponentially stabilizing the SLPPDE system (1) in the sense of $\|\cdot\|_\infty$. The corresponding control gains $\mathbf{k}_{vi}$, $v \in \mathcal{M}$, $i \in \mathbb{S}$ are constructed as (26) from feasible solutions to the LMIs (24), which are directly solved by *feasp* solver in Matlab's LMI Control Toolbox [47].

Based on Theorem 1 and Lemma 4, we will give a less conservative LMI-based design for static sampled-data fuzzy controller (6).

Theorem 2: For given constants $L > 0$, $T_h > 0$, an integer $m > 0$, and parameters $\tilde{x}_{2v-1}$, $\tilde{x}_{2v}$, $v \in \mathcal{M}$, $x_2, x_3, \dots, x_m$, consider the SLPPDE system (1) and the T-S fuzzy PDE model (4). If there exist matrices $0 < \mathbf{X}_1 \in \mathbb{R}^{n \times n}$, $0 < \mathbf{X}_2 \in \mathbb{R}^{n \times n}$, $0 < \mathbf{X}_3 \in \mathbb{R}^{n \times n}$, vectors $\mathbf{z}_{vj} \in \mathbb{R}^n$, and matrices $0 > \Pi_{vij} \in \mathbb{R}^{5n \times 5n}$, $0 > \mathbf{N}_{vij} \in \mathbb{R}^{5n \times 5n}$, $\Sigma_{vij} = \Sigma_{vji}^T \in \mathbb{R}^{5n \times 5n}$, $\Sigma_{vi(j+r)} = \Sigma_{v(j+r)i}^T \in \mathbb{R}^{5n \times 5n}$, $v \in \mathcal{M}$, $i, j \in \mathbb{S}$ such that the following LMIs hold:

$$\bar{\mathbf{X}}_{vij} + \bar{\mathbf{X}}_{vji} \leq \Sigma_{vij} + \Sigma_{vji}, \quad v \in \mathcal{M}, \quad i, j \in \mathbb{S} \quad (27)$$

$$\begin{aligned} \Phi_{vij} - 2\bar{\mathbf{X}}_{vij} - \sum_{k=1}^r (\bar{\mathbf{X}}_{vik}^+ + \bar{\mathbf{X}}_{vkj}^+) \\ \leq \Sigma_{vi(j+r)} + \Sigma_{v(j+r)i}, \quad v \in \mathcal{M}, \quad i, j \in \mathbb{S} \end{aligned} \quad (28)$$

$$\begin{bmatrix} \mathbf{U}_{v11} & \mathbf{U}_{v12} \\ * & \mathbf{U}_{v11} \end{bmatrix} < 0, \quad v \in \mathcal{M} \quad (29)$$

where $\bar{\mathbf{X}}_{vij} \triangleq \Pi_{vij} - \mathbf{N}_{vij}$, $\bar{\mathbf{X}}_{vij}^+ \triangleq \Pi_{vij} + \mathbf{N}_{vij}$, Φ_{vij} (given in Theorem 1), $v \in \mathcal{M}$, $i, j \in \mathbb{S}$,

$$\mathbf{U}_{v11} \triangleq \begin{bmatrix} \Sigma_{v11} & \cdots & \Sigma_{v1r} \\ \vdots & \ddots & \vdots \\ \Sigma_{vr1} & \cdots & \Sigma_{vrr} \end{bmatrix}, \quad v \in \mathcal{M},$$

$$\mathbf{U}_{v12} \triangleq \begin{bmatrix} \Sigma_{v1(r+1)} & \cdots & \Sigma_{v1(2r)} \\ \vdots & \ddots & \vdots \\ \Sigma_{vr(r+1)} & \cdots & \Sigma_{vr(2r)} \end{bmatrix}, \quad v \in \mathcal{M},$$

then there exists a static sampled-data fuzzy controller (6) for the SLPPDE system (1) guaranteeing exponential stability of the resulting closed-loop system in the sense of $\|\cdot\|_\infty$, where the control gains $\mathbf{k}_{vj}$, $v \in \mathcal{M}$, $j \in \mathbb{S}$ are given by (26).

Proof: The proof of this theorem is easily done with the aid of Theorem 1 and Lemma 4 and thus omitted. ■

By Lemma 4, the exponential stabilization ability of static sampled-data fuzzy controller (6) designed by Theorem 2 is enhanced, which will be verified by Example 1 in Section IV.

Remark 3: The main results (i.e., Theorems 1 and 2) are also applicable for the case of mixed Neumann-Dirichlet boundary conditions $\mathbf{y}_x(x, t)|_{x=0} = \mathbf{y}(L, t) = 0$. This is because both the inequality (11) (it requires $\mathbf{y}_x(x, t)|_{x=0} = 0$ or

$y_x(x, t)|_{x=L} = 0$) in Lemma 1 and the vector-valued Agmon's inequality (it requires $y(0, t) = 0$ or $y(L, t) = 0$) in Lemma 3 are employed in the development of these main results.

Remark 4: By Remark 1, the main results (i.e., Theorems 1 and 2) are also directly employed to solve static sampled-data fuzzy control design for the SLPPDE system (1) in the sense of $\|\cdot\|_2$. In this situation, these main results are also applicable for the case of Neumann boundary conditions $y_x(x, t)|_{x=0} = y_x(x, t)|_{x=L} = 0$ due to that they can be derived in the absence of the vector-valued Agmon's inequality given in Lemma 3.

Remark 5: Although the main results (i.e., Theorems 1 and 2) only address the case of single input for each actuator's active area (see Eq. (2)), they are also revised for the case of multi-input for this active area. On the other hand, by using the technique of spatial differential linear matrix inequality (SDLMI) reported in [41] and [42], the main results of this paper can be revised to address the problem of sampled-data fuzzy control design of the SLPPDE system (1) where $f(y(x, t))$ is replaced by $f(y(x, t), x)$.

Remark 6: In comparison to most recent results in [44]-[46] for fuzzy control design of nonlinear DPSs whose control input is continuous in time, one of the main differences of this paper lies in that fuzzy control design is developed in this paper for semi-linear parabolic PDE systems, where the control input is sampled-data and discontinuous in time. Apart from the sampled-data case, another main difference is that by a variation of vector-valued Poincaré-Wirtinger inequality in 1D space (i.e., Lemma 2) and a vector-valued Agmon's inequality (i.e., Lemma 3), the suggested fuzzy control design proposed in this paper is addressed in the sense of $\|\cdot\|_\infty$, which is stronger than the control design results in [44]-[46] discussed in the sense of $\|\cdot\|_2$ by Remark 1. Moreover, the implementation of the proposed fuzzy controller of this paper requires the sensors only active over some partial areas of the spatial domain, while the sensors are required to be distributed continuously over the entire spatial domain for the implementation of the fuzzy controller proposed in [44] and located at some specified points of the spatial domain for the fuzzy controllers reported in [45] and [46].

IV. NUMERICAL SIMULATION

To illustrate the satisfactory and better performance of the proposed design method, this section considers two semi-linear parabolic PDE systems: the first one is a scalar one subject to mixed Dirichlet-Neumann boundary conditions and the second one is a multi-variable one subject to mixed Neumann-Dirichlet boundary conditions.

Example 1: Consider a semi-linear parabolic PDE system:

$$\begin{cases} y_t(x, t) = y_{xx}(x, t) + f(y(x, t)) + \mathbf{g}^T(x)\mathbf{u}(t) \\ y(0, t) = y_x(x, t)|_{x=1} = 0, \quad y(x, 0) = y_0(x), \end{cases} \quad (30)$$

where $y(x, t)$ is state, $\mathbf{u}(t) \triangleq [u_1(t) \ u_2(t)]^T \in \mathbb{R}^2$ is manipulated control input, $\mathbf{g}(x) \triangleq [g_1(x) \ g_2(x)]^T$ is control influence function describing distribution of piecewise actuators. Set $f(y(x, t)) = 3 \sin(y(x, t))$, $g_1(x) = \begin{cases} 1 & x \in [0, 0.5) \\ 0 & \text{otherwise} \end{cases}$, $g_2(x) = \begin{cases} 2 & x \in [0.5, 1) \\ 0 & \text{otherwise} \end{cases}$, $\mathbf{u}(t) = 0$ and $y_0(x) =$

Fig. 2: Open-loop numerical simulation results: (a) profile of evolution of $y(\cdot, t)$ and (b) trajectory of $\|y(\cdot, t)\|_\infty$

$0.5 \sin(0.5\pi x)$, $x \in [0, 1]$, Fig. 2 provides open-loop numerical simulation results: profile of evolution of $y(\cdot, t)$ and trajectory of $\|y(\cdot, t)\|_\infty$. It is obvious from Fig. 2 that the equilibrium profile $y(\cdot, t) = 0$ of the open-loop system of (30) is unstable in the sense of $\|\cdot\|_\infty$.

According to Example 1 in [46], when $y(x, t) \in (-\pi, \pi)$, $x \in [0, 1]$, the semi-linear system (30) is exactly represented by the following T-S fuzzy PDE system of two rules:

Plant Rule 1:

IF $y(x, t)$ is “about 0”, **THEN**

$$\begin{cases} y_t(x, t) = y_{xx}(x, t) + a_1 y(x, t) + \mathbf{g}^T(x)\mathbf{u}(t) \\ y(0, t) = y_x(x, t)|_{x=1} = 0, \quad y(x, 0) = y_0(x) \end{cases}$$

Plant Rule 2:

IF $y(x, t)$ is “about $-\pi$ or π ”, **THEN**

$$\begin{cases} y_t(x, t) = y_{xx}(x, t) + a_2 y(x, t) + \mathbf{g}^T(x)\mathbf{u}(t) \\ y(0, t) = y_x(x, t)|_{x=1} = 0, \quad y(x, 0) = y_0(x) \end{cases}$$

where $a_1 = 3$ and $a_2 = 3\varpi$. The overall fuzzy PDE system is written as

$$\begin{cases} y_t(x, t) = y_{xx}(x, t) + \sum_{i=1}^2 h_i(y(x, t)) a_i y(x, t) + \mathbf{g}^T(x)\mathbf{u}(t) \\ y(0, t) = y_x(x, t)|_{x=1} = 0, \quad y(x, 0) = y_0(x), \end{cases} \quad (31)$$

where

$$\begin{aligned} h_1(y(x, t)) &= \begin{cases} \frac{\sin(y(x, t)) - \varpi y(x, t)}{(1 - \varpi)y(x, t)} & y(x, t) \neq 0 \\ 1 & y(x, t) = 0 \end{cases} \\ h_2(y(x, t)) &= 1 - h_1(y(x, t)), \end{aligned} \quad (32)$$

with $\varpi \triangleq 0.01/\pi$.

Set $\tilde{x}_1 = 0.2$, $\tilde{x}_2 = 0.3$, $\tilde{x}_3 = 0.7$, and $\tilde{x}_4 = 0.8$, we thus have $\varphi_1 = \varphi_2 = 0.09$. Let $T_h = 0.1360$, $\Theta = 1$, $\mathbf{g}_1 = 1$, $\mathbf{g}_2 = 2$, and $\mathbf{A}_i = a_i$, $i \in \{1, 2\}$. By Theorem 2, solving the LMIs (27)-(29) by *feasp* solver [47] and using

Fig. 3: Closed-loop numerical simulation results: (a) profile of evolution of $y(x, t)$, (b) trajectory of $\|y(\cdot, t)\|_\infty$, and (c) trajectory of sampled-data fuzzy control input $\mathbf{u}(t)$

(26), we obtain control parameters k_{vi} , $v \in \{1, 2\}$, $i \in \{1, 2\}$ as $k_{11} = -3.4965$, $k_{12} = -3.4935$, $k_{21} = -1.7482$, and $k_{22} = -1.7467$. Applying the sampled-data fuzzy controller (6) with the aforementioned control parameters to the system (30), the resulting closed-loop numerical simulation results: profile of evolution of $y(x, t)$, trajectory of $\|y(\cdot, t)\|_\infty$, and the corresponding trajectory of the sampled-data fuzzy control input $\mathbf{u}(t)$ are shown in Fig. 3. What is apparent from Fig. 3 is that the suggested sampled-data fuzzy controller (6) stabilizes the system (30) in the sense of $\|\cdot\|_\infty$. On the other hand, by Theorem 1, when $T_h = 0.1360$, it has been verified that the LMIs (24) are infeasible. Setting $T_h = 0.03654$, solving the LMIs (24) and using (26), we can get the control parameters k_{vi} , $v \in \{1, 2\}$, $i \in \{1, 2\}$ as $k_{1i} = -3.2784$ and $k_{2i} = -1.6392$, $i \in \{1, 2\}$. In this situation, the sampled-data fuzzy controller (6) will be reduced to a linear one. Obviously, the exponential stabilization ability of the sampled-data fuzzy controller (6) designed by Theorem 2 is significantly enhanced by Lemma 4 for this example.

Next, we will provide a comparison study for the system (30) between the sampled-data fuzzy controller (6) and a sampled-data fuzzy modal-feedback controller from [39] and [40]. By following the main idea of the control design in [39] and [40], we construct a sampled-data fuzzy modal-feedback

Fig. 4: Closed-loop trajectory of $\|y(\cdot, t)\|_2$ using sampled-data fuzzy controllers (6) and (33)

controller as follows:

$$u_v(t) = \sum_{i=1}^2 h_i(x_s(t_k)) k_{vi} x_s(t_k), \\ v \in \{1, 2\}, t \in [t_k, t_{k+1}), k \in \mathfrak{I}, \quad (33)$$

where $\nu(x) = \sqrt{2} \sin(0.5\pi x)$, $x_s(t) \triangleq \int_0^1 \nu(x) y(x, t) dx$, $k_{11} = -5.1$, $k_{12} = -5.2$, $k_{21} = -2.4$, $k_{22} = -2.3$, and

$$h_1(x_s(t)) = \begin{cases} \frac{\varpi_2 x_s(t) - f(x_s(t))}{(\varpi_2 - \varpi_1) x_s(t)} & x_s(t) \neq 0 \\ \frac{\varpi_2 - 1}{\varpi_2 - \varpi_1} & x_s(t) = 0 \end{cases} \\ h_2(x_s(t)) = 1 - h_1(x_s(t)),$$

with $f(x_s(t)) = \int_0^1 \nu(x) \sin(x_s(t) \nu(x)) dx$, $df(x_s(t)) = \int_0^1 \nu^2(x) \cos(x_s(t) \nu(x)) dx$, $\varpi_1 \triangleq \min_{x_s(t) \in [-0.4, 1.2]} \{df(x_s(t))\}$, and $\varpi_2 \triangleq \max_{x_s(t) \in [-0.4, 1.2]} \{df(x_s(t))\}$.

Applying the sampled-data fuzzy modal-feedback controller (33) to the system (30), the closed-loop trajectory of $\|y(\cdot, t)\|_2$ is shown in Fig. 4. According to Remark 4, the sampled-data fuzzy controller (6) can also exponentially stabilize the system (30) in the sense of $\|y(\cdot, t)\|_2$. The closed-loop trajectory of $\|y(\cdot, t)\|_2$ is also given in Fig. 4 for the system (30) driven by the sampled-data fuzzy controller (6). It is clear from Fig. 4 that in comparison to the sampled-data fuzzy modal-feedback controller from [39] and [40], the suggested sampled-data fuzzy controller (6) provides a better control performance.

To further illustrate the improvement of Lemma 4, we provide a numerical comparison among Lemma 2 in [45], Lemma 2 in [46] and Lemma 4. It is easily observed from the fuzzy membership functions $h_1(y(x, t))$ and $h_2(y(x, t))$ defined in (32) that Lemma 2 in [45] cannot be utilized to reduce the conservativeness of the fuzzy control design due to that $h_2(y(x, t)) = 0$ when $y(x, t) = 0$. By Theorem 1 and Lemma 2 in [46] with $\varsigma_{v,i} = 1.2$, $v, i \in \{1, 2\}$, it has been verified that the corresponding LMI conditions are infeasible for $T_h = 0.1360$. That is, the improvement of Lemma 4 is better than that of Lemma 2 in [46] with $\varsigma_{v,i} = 1.2$, $v, i \in \{1, 2\}$.

Example 2: Consider piecewise sampled-data control of multi-variable parabolic PDE systems with piecewise control

Fig. 5: Open-loop numerical simulation results: (a) profiles of evolution of $y_1(\cdot, t)$ and $y_2(\cdot, t)$, and (b) trajectories of $\|y_1(\cdot, t)\|_\infty$ and $\|y_2(\cdot, t)\|_\infty$

inputs and mixed Neumann-Dirichlet boundary conditions

$$\begin{cases} y_{1,t}(x, t) = y_{1,xx}(x, t) + 3y_1(x, t) \\ \quad - y_2(x, t) - y_1^3(x, t) + \bar{\mathbf{g}}^T(x)\mathbf{u}(t), \\ y_{2,t}(x, t) = y_{2,xx}(x, t) + 0.45y_1(x, t) - 0.1y_2(x, t) \\ y_{i,x}(x, t)|_{x=0} = y_i(1, t) = 0, \quad i \in \{1, 2\} \\ y_i(x, 0) = y_{i,0}(x), \quad i \in \{1, 2\} \end{cases} \quad (34)$$

where $y_i(x, t)$, $i \in \{1, 2\}$ are state variables, $\mathbf{u}(t) \triangleq [u_1(t) \ u_2(t)]^T \in \mathbb{R}^2$ is manipulated input, $\bar{\mathbf{g}}(x) \triangleq [\bar{g}_1(x) \ \bar{g}_2(x)]^T$ is influence function describing distribution of piecewise actuators, where $\bar{g}_1(x) = \begin{cases} 1 & x \in [0, 0.5) \\ 0 & \text{otherwise} \end{cases}$

and $\bar{g}_2(x) = \begin{cases} 2 & x \in [0.5, 1) \\ 0 & \text{otherwise} \end{cases}$. $y_{i,0}(x)$, $i \in \{1, 2\}$ are initial conditions. The PDE in the system (34) is a FitzHugh-Nagumo (FHN) equation that is widely applied to model wavy behavior in excitable media in biology [49] and chemistry [50], where the state variables $y_i(x, t)$, $i \in \{1, 2\}$ are usually termed as the concentrations of “activator” and “inhibitor”, respectively. The initial conditions in (34) are assumed to be $y_{1,0}(x) = 0.4 \cos(0.5\pi x)$ and $y_{2,0}(x) = 0.2 \cos(0.5\pi x) \sin(\pi x)$. Fig. 5 shows open-loop profiles of evolution of $y_1(x, t)$ and $y_2(x, t)$, and open-loop trajectories of $\|y_1(\cdot, t)\|_\infty$ and $\|y_2(\cdot, t)\|_\infty$. It is easily seen from Fig. 5 that operating steady profiles $y_i(\cdot, t) = 0$, $i \in \{1, 2\}$ of the system (34) are unstable in the sense of $\|\cdot\|_\infty$.

Let $\mathbf{y}(x, t) \triangleq [y_1(x, t) \ y_2(x, t)]^T$, $\mathbf{g}_1 = [1 \ 0]^T$, and $\mathbf{g}_2 = [2 \ 0]^T$, the system (34) is rewritten as the form (1), where $\Theta = \mathbf{I}$, $m = 2$, $x_1 = 0$, $x_2 = 0.5$, $x_3 = 1$, and

$$\mathbf{f}(\mathbf{y}(x, t)) = \begin{bmatrix} 3y_1(x, t) - y_2(x, t) - y_1^3(x, t) \\ 0.45y_1(x, t) - 0.1y_2(x, t) \end{bmatrix}.$$

It is clear from Fig. 5 that $y_1(\cdot, t) \in [-0.6, 0.6]$, $t \geq 0$. The operating domain $\mathfrak{S}$ is chosen as $\mathfrak{S} = [-1.2, 1.2]$. By the T-S fuzzy PDE modeling approach in Section IV [41], when $y_1(\cdot, t) \in \mathfrak{S}$, the system (34) can be exactly represented by the following T-S fuzzy PDE system of two rules:

Plant Rule 1:

IF $\xi(x, t)$ is “Big”, **THEN**

$$\begin{cases} \mathbf{y}_t(x, t) = \Theta \mathbf{y}_{xx}(x, t) + \mathbf{A}_1 \mathbf{y}(x, t) + \mathbf{G}(x) \mathbf{u}(t) \\ \mathbf{y}_x(x, t)|_{x=0} = \mathbf{y}(1, t) = 0, \quad \mathbf{y}(x, 0) = \mathbf{y}_0(x) \end{cases}$$

Plant Rule 2:

IF $\xi(x, t)$ is “Small”, **THEN**

$$\begin{cases} \mathbf{y}_t(x, t) = \Theta \mathbf{y}_{xx}(x, t) + \mathbf{A}_2 \mathbf{y}(x, t) + \mathbf{G}(x) \mathbf{u}(t) \\ \mathbf{y}_x(x, t)|_{x=0} = \mathbf{y}(1, t) = 0, \quad \mathbf{y}(x, 0) = \mathbf{y}_0(x) \end{cases}$$

where $\xi(x, t) = y_1^2(x, t)$,

$$\mathbf{A}_1 = \begin{bmatrix} 3 - \vartheta & -1 \\ 0.45 & -0.1 \end{bmatrix} \quad \text{and} \quad \mathbf{A}_2 = \begin{bmatrix} 3 & -1 \\ 0.45 & -0.1 \end{bmatrix}$$

with $\vartheta \triangleq \max_{y_1(\cdot, t) \in \mathfrak{S}} y_1^2(\cdot, t) = 1.44$. The overall fuzzy model is written as follows:

$$\begin{cases} \mathbf{y}_t(x, t) = \Theta \mathbf{y}_{xx}(x, t) \\ \quad + \sum_{i=1}^2 h_i(\xi(x, t)) \mathbf{A}_i \mathbf{y}(x, t) + \mathbf{G}(x) \mathbf{u}(t) \\ \mathbf{y}_x(x, t)|_{x=0} = \mathbf{y}(1, t) = 0, \quad \mathbf{y}(x, 0) = \mathbf{y}_0(x), \end{cases}$$

where the fuzzy membership functions $h_1(\xi(x, t))$ and $h_2(\xi(x, t))$ are chosen as

$$h_1(\xi(x, t)) = \vartheta^{-1} \xi(x, t) \quad \text{and} \quad h_2(\xi(x, t)) = 1 - h_1(\xi(x, t)).$$

For more details, please refer to [41].

Set $\tilde{x}_1 = 0.2$, $\tilde{x}_2 = 0.3$, $\tilde{x}_3 = 0.7$, and $\tilde{x}_4 = 0.8$, we thus have $\varphi_1 = \varphi_2 = 0.09$. Let $T_h = 0.14$. By solving LMIs (27)-(29) and using (26), the control gain parameters $\mathbf{k}_{vi}^T$, $v \in \{1, 2\}$, $i \in \{1, 2\}$ are given as:

$$\begin{bmatrix} \mathbf{k}_{11}^T \\ \mathbf{k}_{21}^T \end{bmatrix} = \begin{bmatrix} -3.4059 & 0.2105 \\ -1.7029 & 0.1053 \end{bmatrix},$$

and

$$\begin{bmatrix} \mathbf{k}_{12}^T \\ \mathbf{k}_{22}^T \end{bmatrix} = \begin{bmatrix} -3.4073 & 0.2104 \\ -1.7036 & 0.1052 \end{bmatrix}.$$

By applying the suggested sampled-data fuzzy controller (6) with above control gain parameters, closed-loop numerical simulation results: profiles of evolution of $y_1(x, t)$ and $y_2(x, t)$, trajectories of $\|y_1(\cdot, t)\|_\infty$ and $\|y_2(\cdot, t)\|_\infty$, and trajectory of sampled-data fuzzy control input $\mathbf{u}(t)$ are shown in Fig. 6. What is apparent from Fig. 6 is that the suggested sampled-data fuzzy controller (6) can stabilize the system (34) in the sense of $\|\cdot\|_\infty$.

V. CONCLUSIONS

This paper has proposed an LMI-based static sampled-data fuzzy local state feedback control design for a class of SLPPDE systems in the sense of $\|\cdot\|_\infty$. In the proposed design development, three types of spatial integral inequalities are utilized, i.e., vector-valued Wirtinger’s inequalities, a variation of vector-valued Poincaré–Wirtinger inequality in 1D spatial domain, and a vector-valued Agmon’s inequality, respectively.

Fig. 6: Closed-loop numerical simulation results: (a) profiles of evolution of $y_1(x, t)$ and $y_2(x, t)$, (b) trajectories of $\|y_1(\cdot, t)\|_\infty$ and $\|y_2(\cdot, t)\|_\infty$, (c) trajectory of sampled-data fuzzy control input $u(t)$

It has been shown by constructing an appropriate Lyapunov–Krasovskii functional candidate and utilizing integration by parts and these three integral inequalities that the suggested sampled-data fuzzy controller exponentially stabilizes the PDE system in the sense of $\|\cdot\|_\infty$, if the LMI-based sufficient conditions are satisfied. Moreover, the exponential stabilization ability of the suggested sampled-data fuzzy controller is enhanced by proposing an LMI relaxation technique. Finally, the satisfactory and better performance of the suggested sampled-data fuzzy controller are demonstrated by numerical simulation results of two examples.

Note that main results of this paper are developed by Lemma 2. To utilize this lemma, we assume $0 < \Theta \in \mathbb{R}^{n \times n}$ in the SLPPDE system (1) such that the inequality $[P_1 \Theta +] > 0$ (see Eq. (49)) is fulfilled for any $P_1 > 0$. Indeed, this assumption limits the scope of the application of the main results. In the further study, we will study the problem of sampled-data fuzzy control design for the SLPPDE system (1) with any general matrix $\Theta \in \mathbb{R}^{n \times n}$.

ACKNOWLEDGMENT

The authors also gratefully acknowledge the helpful comments and suggestions of the Associate Editor and anonymous reviewers, which have improved the presentation of this paper.

APPENDIX A PROOF OF LEMMA 2

Proof: In the light of the eigendecomposition for the matrix $S \geq 0$, we get

$$S = U \Lambda U^T, \quad (35)$$

where U is an orthogonal matrix and $0 \leq \Lambda \triangleq \text{diag}\{\lambda_1, \lambda_2, \dots, \lambda_n\}$ whose entries are eigenvalues of S . By using (35) and setting

$$\tilde{z}(x, t) \triangleq U^T y(x, t) \text{ and } \tilde{z}_l(t) \triangleq U^T y_l(t), \quad (36)$$

where $\tilde{z}(x, t) = [\tilde{z}_1(x, t) \ \tilde{z}_2(x, t) \ \dots \ \tilde{z}_n(x, t)]^T$ and $\tilde{z}_l(t) = [\tilde{z}_{l,1}(t) \ \tilde{z}_{l,2}(t) \ \dots \ \tilde{z}_{l,n}(t)]^T$, with

$$\tilde{z}_{l,i}(t) = \frac{1}{l_2 - l_1} \int_{l_1}^{l_2} \tilde{z}_i(x, t) dx, \quad i \in \{1, 2, \dots, n\}. \quad (37)$$

By first mean value theorem for definite integrals [51], for any $y(\cdot, t) \in \mathcal{H}_n^1((0, L))$ ($\tilde{z}(\cdot, t) \in \mathcal{H}_n^1((0, L))$, obviously), there exists scalars $\bar{l}_{i,t} \in [l_1, l_2]$, $i \in \{1, 2, \dots, n\}$ such that

$$\int_{l_1}^{l_2} \tilde{z}_i(x, t) dx = (l_2 - l_1) \tilde{z}_i(\bar{l}_{i,t}, t), \quad i \in \{1, 2, \dots, n\} \quad (38)$$

where $\tilde{z}_i(\bar{l}_{i,t}, t)$ is called *mean value* of $\tilde{z}_i(x, t)$ on $[l_1, l_2]$. The scalars $\bar{l}_{i,t}$, $i \in \{1, 2, \dots, n\}$ such that the Eq. (38) is fulfilled are not unique and their accurate value is unknown in general. From (36)-(38), we get

$$\begin{aligned} & \int_0^L (y(x, t) - y_l(t))^T S (y(x, t) - y_l(t)) dx \\ &= \int_0^L (\tilde{z}(x, t) - \tilde{z}_l(t))^T \Lambda (\tilde{z}(x, t) - \tilde{z}_l(t)) dx \\ &= \sum_{i=1}^n \int_0^L \lambda_i (\tilde{z}_i(x, t) - \tilde{z}_i(\bar{l}_{i,t}, t))^2 dx, \end{aligned} \quad (39)$$

For each $i \in \{1, 2, \dots, n\}$, by applying extended scalar Wirtinger's inequality [48] and considering $\bar{l}_{i,t} \in [l_1, l_2]$, the following inequality is easily derived

$$\begin{aligned} & \int_0^L (\tilde{z}_i(x, t) - \tilde{z}_i(\bar{l}_{i,t}, t))^2 dx \\ & \leq \frac{4 \max\{\bar{l}_{i,t}^2, (L - \bar{l}_{i,t})^2\}}{\pi^2} \int_0^L \tilde{z}_{i,x}^2(x, t) dx \\ & \leq \frac{4\phi}{\pi^2} \int_0^L \tilde{z}_{i,x}^2(x, t) dx. \end{aligned} \quad (40)$$

It is deduced from (36), (40) and $\lambda_i \geq 0$, $i \in \{1, 2, \dots, n\}$ that

$$\begin{aligned} & \sum_{i=1}^n \int_0^L \lambda_i (\tilde{z}_i(x, t) - \tilde{z}_i(\bar{l}_{i,t}, t))^2 dx \\ & \leq \frac{4\phi}{\pi^2} \sum_{i=1}^n \int_0^L \lambda_i \tilde{z}_{i,x}^2(x, t) dx \\ & = \frac{4\phi}{\pi^2} \int_0^L \tilde{\mathbf{z}}_x^T(x, t) \mathbf{\Lambda} \tilde{\mathbf{z}}_x(x, t) dx \\ & = \frac{4\phi}{\pi^2} \int_0^L \mathbf{y}_x^T(x, t) \mathbf{S} \mathbf{y}_x(x, t) dx. \end{aligned} \quad (41)$$

The inequality (12) is directly derived from (39) and (41). ■

APPENDIX B PROOF OF LEMMA 3

Proof: Using integration by parts and considering $\mathbf{y}(0, t) = 0$, we have for any $t \geq 0$

$$\begin{aligned} \int_0^x \mathbf{y}^T(\varsigma, t) \mathbf{y}_\varsigma(\varsigma, t) d\varsigma &= \mathbf{y}^T(x, t) \mathbf{y}(x, t) \\ &\quad - \int_0^x \mathbf{y}_\varsigma^T(\varsigma, t) \mathbf{y}(\varsigma, t) d\varsigma. \end{aligned} \quad (42)$$

In the light of Cauchy–Schwarz inequality, the expression (42) is rewritten as for any $x \in [0, L]$ and $t \geq 0$

$$\begin{aligned} \mathbf{y}^T(x, t) \mathbf{y}(x, t) &= 2 \int_0^x \mathbf{y}^T(\varsigma, t) \mathbf{y}_\varsigma(\varsigma, t) d\varsigma \\ &\leq \sqrt{\int_0^x \|\mathbf{y}(\varsigma, t)\|^2 d\varsigma} \sqrt{\int_0^x \|\mathbf{y}_\varsigma(\varsigma, t)\|^2 d\varsigma} \\ &\leq \|\mathbf{y}(\cdot, t)\|_2 \|\mathbf{y}_x(\cdot, t)\|_2. \end{aligned} \quad (43)$$

As it is fulfilled for any $x \in [0, L]$, the inequality (43) is further written as $\max_{x \in [0, L]} \|\mathbf{y}(x, t)\|^2 \leq 2 \|\mathbf{y}(\cdot, t)\|_2 \|\mathbf{y}_x(\cdot, t)\|_2$, $t \geq 0$, which implies the first inequality in (13) for the case of $\mathbf{y}(0, t) = 0$. By using the triangle inequality and considering the definition of $\|\cdot\|_{\mathcal{H}_n^1((0, L))}$, we have $2 \|\mathbf{y}(\cdot, t)\|_2 \|\mathbf{y}_x(\cdot, t)\|_2 \leq \|\mathbf{y}(\cdot, t)\|_{\mathcal{H}_n^1((0, L))}^2$, $t \geq 0$, which is the second inequality in (13). The inequality (13) is easily derived for the case of $\mathbf{y}(L, t) = 0$ in a similar manner. ■

APPENDIX C PROOF OF THEOREM 1

Proof: Assume that the LMIs (24) are fulfilled for matrices $0 < \mathbf{X}_1 \in \mathbb{R}^{n \times n}$, $0 < \mathbf{X}_2 \in \mathbb{R}^{n \times n}$, $0 < \mathbf{X}_3 \in \mathbb{R}^{n \times n}$ and vectors $\mathbf{z}_{vj} \in \mathbb{R}^n$, $v \in \mathcal{M}$, $j \in \mathbb{S}$. By applying integration by parts and considering the boundary conditions in (1), we get for any $\mathbf{P}_1$, $\mathbf{P}_2$ and $\mathbf{\Theta}$

$$\int_0^L \mathbf{y}^T(x, t) \mathbf{P}_1 \mathbf{\Theta} \mathbf{y}_{xx}(x, t) dx = - \int_0^L \mathbf{y}_x^T(x, t) \mathbf{P}_1 \mathbf{\Theta} \mathbf{y}_x(x, t) dx, \quad (44)$$

$$\int_0^L \mathbf{y}_x^T(x, t) \mathbf{P}_2 \mathbf{y}_{xt}(x, t) dx = - \int_0^L \mathbf{y}_{xx}^T(x, t) \mathbf{P}_2 \mathbf{y}_t(x, t) dx. \quad (45)$$

Define

$$\tilde{\mathbf{y}}_v(x, t) \triangleq \mathbf{y}(x, t) - \frac{1}{\Delta \tilde{x}_v} \int_{\tilde{x}_{2v-1}}^{\tilde{x}_{2v}} \mathbf{y}(x, t) dx, \quad v \in \mathcal{M}, \quad (46)$$

$$\bar{\mathbf{y}}_v(t, t_k) \triangleq \int_{\tilde{x}_{2v-1}}^{\tilde{x}_{2v}} \int_{t_k}^t \mathbf{y}_s(x, s) ds dx, \quad v \in \mathcal{M}, \quad k \in \wp, \quad (47)$$

where $\Delta \tilde{x}_v$, $v \in \mathcal{M}$ are defined by (7). Using (44) and (47), the time derivative of $V_1(t)$ in (19) along the solution to the system (9), we get for any $t \in [t_k, t_{k+1})$, $k \in \wp$

$$\begin{aligned} \dot{V}_1(t) &= - \int_0^L \mathbf{y}_x^T(x, t) [\mathbf{P}_1 \mathbf{\Theta} + *] \mathbf{y}_x(x, t) dx \\ &\quad + \int_0^L \sum_{i=1}^r h_i(\xi(x, t)) \mathbf{y}^T(x, t) [\mathbf{P}_1 \mathbf{A}_i + *] \mathbf{y}(x, t) dx \\ &\quad + 2 \sum_{v=1}^m \int_{x_v}^{x_{v+1}} \mathbf{y}^T(x, t) \mathbf{P}_1 \mathbf{g}_v dx \\ &\quad \times \sum_{j=1}^r h_{vj}(t_k) \frac{\mathbf{k}_{vj}^T}{\Delta \tilde{x}_v} \int_{\tilde{x}_{2v-1}}^{\tilde{x}_{2v}} \mathbf{y}(x, t) dx \\ &\quad - 2 \sum_{v=1}^m \int_{x_v}^{x_{v+1}} \mathbf{y}^T(x, t) \mathbf{P}_1 \mathbf{g}_v dx \\ &\quad \times \sum_{j=1}^r h_{vj}(t_k) \frac{\mathbf{k}_{vj}^T}{\Delta \tilde{x}_v} \bar{\mathbf{y}}_v(t, t_k). \end{aligned} \quad (48)$$

It can be obtained from $\mathbf{P}_1 > 0$ and $\mathbf{\Theta} > 0$ that

$$\mathbf{\Psi} \triangleq [\mathbf{P}_1 \mathbf{\Theta} + *] > 0. \quad (49)$$

By using the inequality (12) in Lemma 2 and considering (49) and $[\tilde{x}_{2v-1}, \tilde{x}_{2v}] \subset [x_v, x_{v+1}]$, we get for any $v \in \mathcal{M}$,

$$\begin{aligned} & - \int_{x_v}^{x_{v+1}} \mathbf{y}_x^T(x, t) \mathbf{\Psi} \mathbf{y}_x(x, t) dx \\ & \leq - \frac{\pi^2}{4\varphi_v} \int_{x_v}^{x_{v+1}} \tilde{\mathbf{y}}_v^T(x, t) \mathbf{\Psi} \tilde{\mathbf{y}}_v(x, t) dx, \end{aligned} \quad (50)$$

where φ_v , $v \in \mathcal{M}$ are defined by (25). Substituting (50) into (48) and considering $\bigcup_{v=1}^m [x_v, x_{v+1}] = [0, L]$, we get

$$\begin{aligned} \dot{V}_1(t) &\leq \sum_{v=1}^m \int_{x_v}^{x_{v+1}} \sum_{i=1}^r \sum_{j=1}^r h_i(\xi(x, t)) h_{vj}(t) \\ &\quad \times \hat{\mathbf{y}}_v^T(x, t) \bar{\mathbf{\Psi}}_{1vij} \hat{\mathbf{y}}_v(x, t) dx \\ &\quad - 2 \sum_{v=1}^m \int_{x_v}^{x_{v+1}} \mathbf{y}^T(x, t) \mathbf{P}_1 \mathbf{g}_v dx \\ &\quad \times \sum_{j=1}^r h_{vj}(t_k) \frac{\mathbf{k}_{vj}^T}{\Delta \tilde{x}_v} \bar{\mathbf{y}}_v(t, t_k), \end{aligned} \quad (51)$$

where $\hat{\mathbf{y}}_v(x, t) \triangleq [\mathbf{y}^T(x, t) \quad (\Delta \tilde{x}_v)^{-1} \int_{\tilde{x}_{2v-1}}^{\tilde{x}_{2v}} \mathbf{y}^T(x, t) dx]^T$ and

$$\bar{\mathbf{\Psi}}_{1vij} \triangleq \begin{bmatrix} [\mathbf{P}_1 \mathbf{A}_i + *] - \frac{\pi^2}{4\varphi_v} \mathbf{\Psi} & \frac{\pi^2}{4\varphi_v} \mathbf{\Psi} + \mathbf{P}_1 \mathbf{g}_v \mathbf{k}_{vj}^T \\ * & - \frac{\pi^2}{4\varphi_v} \mathbf{\Psi} \end{bmatrix}, \quad v \in \mathcal{M}, \quad i, j \in \mathbb{S}.$$

Similarly, using (45), (47), and $\bigcup_{v=1}^m [x_v, x_{v+1}] = [0, L]$, for $t \in [t_k, t_{k+1})$, $k \in \wp$, the time derivative of $V_2(t)$ defined in (20) along the solution to the system (9) is given as follows:

$$\begin{aligned} \dot{V}_2(t) = & - \int_0^L \mathbf{y}_{xx}^T(x, t) [\mathbf{P}_2 \Theta + *] \mathbf{y}_{xx}(x, t) dx \\ & - 2 \int_0^L \sum_{i=1}^r h_i(\xi(x, t)) \mathbf{y}_{xx}^T(x, t) \mathbf{P}_2 \mathbf{A}_i \mathbf{y}(x, t) dx \\ & - 2 \sum_{v=1}^m \int_{x_v}^{x_{v+1}} \mathbf{y}_{xx}^T(x, t) \mathbf{P}_2 \mathbf{g}_v dx \\ & \times \sum_{j=1}^r h_{vj}(t_k) \frac{\mathbf{k}_{vj}^T}{\Delta \tilde{x}_v} \int_{\tilde{x}_{2v-1}}^{\tilde{x}_{2v}} \mathbf{y}(x, t) dx \\ & + 2 \sum_{v=1}^m \int_{x_v}^{x_{v+1}} \mathbf{y}_{xx}^T(x, t) \mathbf{P}_2 \mathbf{g}_v dx \\ & \times \sum_{j=1}^r h_{vj}(t_k) \frac{\mathbf{k}_{vj}^T}{\Delta \tilde{x}_v} \bar{\mathbf{y}}_v(t, t_k). \end{aligned} \quad (52)$$

Based on the Jensen's inequality [52], we get for any $t \in (t_k, t_{k+1})$, $\mathbf{P}_3 > 0$ and $x \in [0, L]$

$$\begin{aligned} & - \int_{t_k}^t \mathbf{y}_s^T(x, s) \mathbf{P}_3 \mathbf{y}_s(x, s) ds \\ & \leq -T_h^{-1} \int_{t_k}^t \mathbf{y}_s^T(x, s) ds \mathbf{P}_3 \int_{t_k}^t \mathbf{y}_s(x, s) ds. \end{aligned} \quad (53)$$

When $t = t_k$, the inequality (53) is also fulfilled obviously due to that for any $x \in [0, L]$,

$$\begin{aligned} & \lim_{t \rightarrow t_k} \frac{1}{t - t_k} \int_{t_k}^t \mathbf{y}_s^T(x, s) ds \mathbf{P}_3 \int_{t_k}^t \mathbf{y}_s(x, s) ds \\ & = 2 \lim_{t \rightarrow t_k} \int_{t_k}^t \mathbf{y}_s^T(x, s) ds \mathbf{P}_3 \mathbf{y}_t(x, t) = 0. \end{aligned}$$

Utilizing $\bigcup_{v=1}^m [x_v, x_{v+1}] = [0, L]$ and (53), for $t \in [t_k, t_{k+1})$, $k \in \wp$, the time derivative of $V_3(t)$ defined in (21) is obtained as follows

$$\begin{aligned} \dot{V}_3(t) \leq & T_h^2 \sum_{v=1}^m \int_{x_v}^{x_{v+1}} \mathbf{y}_t^T(x, t) \mathbf{P}_3 \mathbf{y}_t(x, t) dx \\ & - \sum_{v=1}^m \int_{x_v}^{x_{v+1}} \int_{t_k}^t \mathbf{y}_s^T(x, s) ds \mathbf{P}_3 \int_{t_k}^t \mathbf{y}_s(x, s) ds dx. \end{aligned} \quad (54)$$

With the help of the Jensen's inequality again and $\bigcup_{v=1}^m [\tilde{x}_{2v-1}, \tilde{x}_{2v}] \subset [0, L] = \bigcup_{v=1}^m [x_v, x_{v+1}]$, we know for any $v \in \mathcal{M}$ and $t \in [t_k, t_{k+1})$, $k \in \wp$

$$\begin{aligned} & - \int_{x_v}^{x_{v+1}} \int_{t_k}^t \mathbf{y}_s^T(x, s) ds \mathbf{P}_3 \int_{t_k}^t \mathbf{y}_s(x, s) ds dx \\ & < -(\Delta \tilde{x}_v)^{-1} \bar{\mathbf{y}}_v^T(t, t_k) \mathbf{P}_3 \bar{\mathbf{y}}_v(t, t_k). \end{aligned} \quad (55)$$

Substitution of (55) into (54), we rewrite (54) as

$$\begin{aligned} \dot{V}_3(t) < & T_h^2 \sum_{v=1}^m \int_{x_v}^{x_{v+1}} \mathbf{y}_t^T(x, t) \mathbf{P}_3 \mathbf{y}_t(x, t) dx \\ & - \sum_{v=1}^m \frac{1}{\Delta x_v \Delta \tilde{x}_v} \int_{x_v}^{x_{v+1}} \bar{\mathbf{y}}_v^T(t, t_k) \mathbf{P}_3 \bar{\mathbf{y}}_v(t, t_k) dx, \end{aligned} \quad (56)$$

where Δx_v , $v \in \mathcal{M}$ are defined by (25).

On the other hand, by considering $\bigcup_{v=1}^m [x_v, x_{v+1}] = [0, L]$, from (9) and (47), it is clearly seen for any $t \geq 0$ that

$$\begin{aligned} 0 = & 2 \sum_{v=1}^m \int_{x_v}^{x_{v+1}} \mathbf{y}_t^T(x, t) \mathbf{P}_1 \Theta \mathbf{y}_{xx}(x, t) dx \\ & + 2 \sum_{v=1}^m \int_{x_v}^{x_{v+1}} \sum_{i=1}^r h_i(\xi(x, t)) \mathbf{y}_t^T(x, t) \mathbf{P}_1 \mathbf{A}_i \mathbf{y}(x, t) dx \\ & + 2 \sum_{v=1}^m \int_{x_v}^{x_{v+1}} \mathbf{y}_t^T(x, t) \mathbf{P}_1 \mathbf{g}_v dx \\ & \times \sum_{j=1}^r h_{vj}(t_k) \frac{\mathbf{k}_{vj}^T}{\Delta \tilde{x}_v} \int_{\tilde{x}_{2v-1}}^{\tilde{x}_{2v}} \mathbf{y}(x, t) dx \\ & - 2 \sum_{v=1}^m \int_{x_v}^{x_{v+1}} \mathbf{y}_t^T(x, t) \mathbf{P}_1 \mathbf{g}_v dx \sum_{j=1}^r h_{vj}(t_k) \frac{\mathbf{k}_{vj}^T}{\Delta \tilde{x}_v} \bar{\mathbf{y}}_v(t, t_k) \\ & - 2 \sum_{v=1}^m \int_{x_v}^{x_{v+1}} \mathbf{y}_t^T(x, t) \mathbf{P}_1 \mathbf{y}_t(x, t) dx. \end{aligned} \quad (57)$$

By using (51), (52), (56), and (57), for $t \in [t_k, t_{k+1})$, $k \in \wp$, the time derivative of $V(t)$ defined in (18) along the solution to the system (9) is given as follows:

$$\begin{aligned} \dot{V}(t) < & \sum_{v=1}^m \int_{x_v}^{x_{v+1}} \sum_{i=1}^r \sum_{j=1}^r h_i(\xi(x, t)) h_{vj}(t) \\ & \times \zeta_v^T(x, t, t_k) \bar{\Psi}_{vij} \zeta_v(x, t, t_k) dx, \end{aligned} \quad (58)$$

where

$$\zeta_v(x, t, t_k) \triangleq \begin{bmatrix} \mathbf{y}_v^T(x, t) & \mathbf{y}_{xx}^T(x, t) & \mathbf{y}_t^T(x, t) & \frac{\bar{\mathbf{y}}_v^T(t, t_k)}{\Delta \tilde{x}_v} \end{bmatrix}^T,$$

$$\bar{\Psi}_{vij} \triangleq \begin{bmatrix} \bar{\Psi}_{1vij} & \bar{\Psi}_{2vij} \\ * & \bar{\Psi}_{3vij} \end{bmatrix}, \quad v \in \mathcal{M}, \quad i, j \in \mathbb{S}$$

in which

$$\begin{aligned} \bar{\Psi}_{2vij} & \triangleq \begin{bmatrix} -\mathbf{A}_i^T \mathbf{P}_2 & \mathbf{A}_i^T \mathbf{P}_1 & -\mathbf{P}_1 \mathbf{g}_v \mathbf{k}_{vj}^T \\ -\mathbf{k}_{vj} \mathbf{g}_v^T \mathbf{P}_2 & \mathbf{k}_{vj} \mathbf{g}_v^T \mathbf{P}_1 & 0 \end{bmatrix}, \\ \bar{\Psi}_{3vij} & \triangleq \begin{bmatrix} -[\mathbf{P}_2 \Theta + *] & \Theta \mathbf{P}_1 & \mathbf{P}_2 \mathbf{g}_v \mathbf{k}_{vj}^T \\ * & T_h^2 \mathbf{P}_3 - 2\mathbf{P}_1 & -\mathbf{P}_1 \mathbf{g}_v \mathbf{k}_{vj}^T \\ * & * & -\frac{\Delta \tilde{x}_v}{\Delta x_v} \mathbf{P}_3 \end{bmatrix}. \end{aligned}$$

Let

$$\begin{aligned} \mathbf{X}_1 & = \mathbf{P}_1^{-1}, \quad \mathbf{X}_2 = \mathbf{P}_2^{-1}, \quad \mathbf{X}_3 = \mathbf{P}_1^{-1} \mathbf{P}_3 \mathbf{P}_1^{-1}, \\ \mathbf{z}_{vj}^T & = \mathbf{k}_{vj}^T \mathbf{X}_1, \quad v \in \mathcal{M}, \quad j \in \mathbb{S}. \end{aligned} \quad (59)$$

By pre- and post-multiplying the LMIs (24) with a block-diagonal matrix $\mathbb{P} \triangleq \text{diag}\{\mathbf{P}_1, \mathbf{P}_1, \mathbf{P}_2, \mathbf{P}_1, \mathbf{P}_1\}$, respectively, and using (59), we obtain

$$\bar{\Psi}_{vij} < 0, \quad v \in \mathcal{M}, \quad i, j \in \mathbb{S}. \quad (60)$$

One can find an appropriate scalar $\rho > 0$ such that

$$\bar{\Psi}_{vij} + \rho \mathbf{I} \leq 0, \quad v \in \mathcal{M}, \quad i, j \in \mathbb{S}. \quad (61)$$

Substituting (61) into (58) and considering (5) and (8), we obtain for $t \in [t_k, t_{k+1})$, $k \in \wp$

$$\begin{aligned} \dot{V}(t) &< -\rho \int_0^L \mathbf{y}^T(x, t) \mathbf{y}(x, t) dx \\ &\quad - \rho \int_0^L \mathbf{y}_{xx}^T(x, t) \mathbf{y}_{xx}(x, t) dx - \rho \int_0^L \mathbf{y}_t^T(x, t) \mathbf{y}_t(x, t) dx. \end{aligned} \quad (62)$$

By applying the inequality (11) in Lemma 1 and considering $\mathbf{y}_x(x, t)|_{x=L} = 0$, we get for any $t > 0$

$$\int_0^L \mathbf{y}_x^T(x, t) \mathbf{y}_x(x, t) dx \leq \frac{4L^2}{\pi^2} \int_0^L \mathbf{y}_{xx}^T(x, t) \mathbf{y}_{xx}(x, t) dx. \quad (63)$$

By following proofs of Theorem 3 in [48] and Theorem 1 in [53], it is easily shown from the inequalities (62) and (63) that for any $t_f \geq 0$ there exists a constant $\kappa > 0$ such that

$$V(t_f) \leq V(0) \exp(-\kappa t_f). \quad (64)$$

By (22) and (23), the inequality (64) is written as $\|\mathbf{y}(\cdot, t_f)\|_{\mathcal{H}_n^1((0, L))}^2 \leq \rho_2 \rho_1^{-1} \|\mathbf{y}_0(\cdot)\|_{\mathcal{H}_n^1((0, L))}^2 \exp(-\kappa t_f)$. Based on Lemma 3, we further get $\|\mathbf{y}(\cdot, t_f)\|_\infty \leq \sqrt{\rho_2 \rho_1^{-1}} \|\mathbf{y}_0(\cdot)\|_{\mathcal{H}_n^1((0, L))} \exp(-0.5\kappa t_f)$. Hence, we conclude from the above inequality and Definition 2 that the system (9) is exponentially stable in the sense of $\|\cdot\|_\infty$. Consider $\mathbb{D} \subset \mathfrak{S}$ and the equivalence between the SLPPDE in (1) and the fuzzy PDE (4) in the operating domain $\mathfrak{S}$, the suggested sampled-data fuzzy controller (6) can locally exponentially stabilize the system (1) in the sense of $\|\cdot\|_\infty$. From (59), we obtain (26). ■

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Jun-Wei Wang received the B. S. degree in mathematics and applied mathematics and the M. S. degree in system theory both from Harbin Engineering University, Harbin, China, in 2007 and 2009, and the Ph. D. degree in control theory and control engineering from Beihang University, Beijing, China, in 2013, respectively.

From March to September in 2012, he was a Research Assistant with the Department of Systems Engineering and Engineering Management, City University of Hong Kong. From September 2013 to

December 2015, He was a Teacher/Postdoctoral Researcher with the School of Automation and Electrical Engineering, University of Science and Technology Beijing. From August 2014 to July 2015, he was a Postdoctoral Researcher in the Department of Electrical and Computer Engineering, Stevens Institute of Technology, New Jersey USA. From February 2016 to February 2018, He was a Postdoctoral Fellow under the Hong Kong Scholars Scheme at City University of Hong Kong. In January 2016, he joined the School of Automation and Electrical Engineering, University of Science and Technology Beijing, where he is currently a Lecturer.

His research interests include robust control and filtering, stochastic systems, time-delay systems, distributed parameter systems, and fuzzy/neural network modeling and control.

Shun-Hung Tsai Shun-Hung Tsai received the B. S. degree from the Department of Industrial Education, National Taiwan Normal University, Taipei, Taiwan, R.O.C. in 1999, and the M. S. and Ph. D. degrees in 2001 and 2007, respectively, from National Cheng Kung University, Tainan, Taiwan, R.O.C., all in electrical engineering.

He was an electrical engineer in Philips Building Semiconductor from 2003 to 2005. From 2005 to 2006, he was a control engineer of China Steel Corporation. From 2006 to 2007, he served as a process integration engineer in United Microelectronics Corporation. In 2008, he was an associate researcher at the Energy and Agile System Development, Metal Industrial Research and Develop Centre (MIRDC), Kaohsiung, Taiwan. Besides, he was an Assistant Professor in the department of electrical engineering, Chang Gung University of Taiwan in 2008. Furthermore, from 2009–2012, he was an Assistant Professor in the Graduate Institute of Automation Technology, National Taipei University of Technology. He is currently an Associate Professor in the Graduate Institute of Automation Technology, National Taipei University of Technology.

His major research interests include nonlinear control, intelligent control, system identification, robotics and quadrotor.

Han-Xiong Li (S'94-M'97-SM'00-F'11) received his B. E. degree in aerospace engineering from the National University of Defense Technology, China in 1982, M. E. degree in electrical engineering from Delft University of Technology, The Netherlands in 1991, and Ph.D. degree in electrical engineering from the University of Auckland, New Zealand in 1997.

He is a professor in the Department of SEEM, City University of Hong Kong. He has a broad experience in both academia and industry. He has

authored 2 books and 7 patents, and published more than 200 SCI journal papers with h-index 37 (web of science). His current research interests include process modeling and control, system intelligence, distributed parameter systems, and battery management system.

Dr. Li serves as Associate Editor for IEEE Transactions on SMC: System, and was associate editor for IEEE Transactions on Cybernetics (2002–2016) and IEEE Transactions on Industrial Electronics (2009–2015). He was awarded the Distinguished Young Scholar (overseas) by the China National Science Foundation in 2004, a Chang Jiang professorship by the Ministry of Education, China in 2006, and a national professorship in China Thousand Talents Program in 2010. He serves as a distinguished expert for Hunan Government and China Federation of Returned Overseas Chinese. He is a Fellow of the IEEE.

Hak-Keung Lam received the B. Eng. (Hons.) and Ph. D. degrees from the Department of Electronic and Information Engineering, The Hong Kong Polytechnic University, Hong Kong, in 1995 and 2000, respectively. During the period of 2000 and 2005, he worked with the Department of Electronic and Information Engineering at The Hong Kong Polytechnic University as Post-Doctoral Fellow and Research Fellow respectively. He joined as a Lecturer at Kings College London in 2005 and is currently a Reader.

His current research interests include intelligent control and computational intelligence. He has served as a program committee member, international advisory board member, invited session chair and publication chair for various international conferences and a reviewer for various books, international journals and international conferences. He is an associate editor for IEEE Transactions on Fuzzy Systems, IEEE Transactions on Circuits and Systems II: Express Briefs, IET Control Theory and Applications, International Journal of Fuzzy Systems and Neurocomputing; and guest editor for a number of international journals. He is also in the editorial board of many other journals. He is an IEEE senior member.

He is a coeditor of two edited volumes: Control of Chaotic Nonlinear Circuits (World Scientific, 2009) and Computational Intelligence and Its Applications (World Scientific, 2012), and author/coauthor of three monographs: Stability Analysis of Fuzzy-Model-Based Control Systems (Springer, 2011), Polynomial Fuzzy Model Based Control Systems (Springer, 2016) and Analysis and Synthesis for Interval Type-2 Fuzzy-Model-Based Systems (Springer, 2016).