



King's Research Portal

DOI:

[10.1109/FUZZ-IEEE.2015.7337896](https://doi.org/10.1109/FUZZ-IEEE.2015.7337896)

Document Version

Peer reviewed version

[Link to publication record in King's Research Portal](#)

Citation for published version (APA):

Tsai, S. H., Jian, S. A., Chen, Y. A., Lam, H. K., & Li, Y. (2015). Delay-dependent stabilization condition for T-S fuzzy neutral systems. In *IEEE International Conference on Fuzzy Systems* (Vol. 2015-November). Article 7337896 Institute of Electrical and Electronics Engineers Inc.. <https://doi.org/10.1109/FUZZ-IEEE.2015.7337896>

Citing this paper

Please note that where the full-text provided on King's Research Portal is the Author Accepted Manuscript or Post-Print version this may differ from the final Published version. If citing, it is advised that you check and use the publisher's definitive version for pagination, volume/issue, and date of publication details. And where the final published version is provided on the Research Portal, if citing you are again advised to check the publisher's website for any subsequent corrections.

General rights

Copyright and moral rights for the publications made accessible in the Research Portal are retained by the authors and/or other copyright owners and it is a condition of accessing publications that users recognize and abide by the legal requirements associated with these rights.

- Users may download and print one copy of any publication from the Research Portal for the purpose of private study or research.
- You may not further distribute the material or use it for any profit-making activity or commercial gain
- You may freely distribute the URL identifying the publication in the Research Portal

Take down policy

If you believe that this document breaches copyright please contact librarypure@kcl.ac.uk providing details, and we will remove access to the work immediately and investigate your claim.

Delay-dependent stabilization condition for T-S fuzzy neutral systems

Shun-Hung Tsai

Institute of Automation Technology
National Taipei University of Technology
Taipei, Taiwan
Email: shtsai@ntut.edu.tw

Siou-An Jian

Institute of Automation Technology
National Taipei University of Technology
Taipei, Taiwan
Email: t102618018@ntut.edu.tw

Yu-An Chen

Institute of Automation Technology
National Taipei University of Technology
Taipei, Taiwan
Email: t102618007@ntut.edu.tw

H. K. Lam

Department of Informatics, Kings College London
London, WC2R 2LS, United Kingdom
Email: hak-keung.lam@kcl.ac.uk

Yuandi Li

Department of Informatics, Kings College London
London, WC2R 2LS, United Kingdom
Email: yuandi.li@kcl.ac.uk

Abstract—In this paper, the stabilization problems for a class of Takagi-Sugeno (T-S) fuzzy neutral systems are explored. Utilizing Pólya's theorem and some homogeneous polynomials techniques, the delay-dependent stabilization condition for T-S fuzzy neutral systems are proposed in terms of a linear matrix inequality (LMI) to guarantee the asymptotic stabilization of T-S fuzzy neutral systems. Lastly, an example is illustrated to demonstrate the effectiveness and applicability of the proposed method.

Keywords—Fuzzy control, Takagi-Sugeno (T-S) fuzzy model, linear matrix inequality (LMI), Pólya's theorem.

I. INTRODUCTION

During the past decades, fuzzy logic control has been widely adopted to analyze nonlinear systems [1], [2]. In addition to fuzzy logic control, a great deal of effort has been devoted to describe a nonlinear model using Takagi-Sugeno (T-S) fuzzy model. Moreover, many fuzzy modelling approaches [3], [4] which are provided to represent nonlinear models as T-S fuzzy model. Through the modeling procedure, T-S fuzzy model can be described by fuzzy IF-THEN rules, which represents local linear input-output relations of a nonlinear model. In addition to fuzzy modeling scheme, the parallel distributed compensation (PDC) technique [5] is adopted to stabilize the overall T-S fuzzy model. For the stabilization condition analysis, Lyapunov direct method [6], [7] is mainly investigated to yield the stabilization conditions of T-S fuzzy model. Furthermore, many studies utilize linear matrix inequalities (LMIs) to find a feasible solution, if available. Furthermore, the solutions can be found via linear matrix inequalities (LMIs) techniques.

Time-delay phenomenon exists in many practical systems, such as chemical engineering systems, robotic arm systems and network systems [8], [9]. In general, the practical systems with time-delay are more complicated than those systems without time-delays. Besides, time-delay may cause instability and reduce the system performance under some situations; therefore, there has been an increasing interest in the stabilization problem for time-delay systems, and a lot of results on these topics have been explored in the literature [10], [11].

In addition, time-delay phenomenon exists in both the state and the derivative of the state in neutral systems. For this reason, stabilization problem for neutral systems has been explored in many studies [12], [13]. For example, in [14], a state matrix decomposition is adopted and a delay-dependent stability condition for fuzzy neutral systems was proposed. A descriptor system approach is adopted for uncertain fuzzy neutral system in [15]. Recently, a polynomial technique has been adopted to reduce the conservatism of the stabilization condition [16]–[18]. Inspired by these works and reference therein, we will explore the delay-dependent stabilization condition for T-S fuzzy neutral systems via polynomial technique.

Following the introduction, the paper is organized as follows. In Section II, a general description of T-S fuzzy neutral system is introduced and the state feedback fuzzy controller is also designed. In Section III, based on homogeneous polynomial technique and Pólya's theorem, a delay-dependent stabilization conditions for T-S fuzzy neutral system is formulated in terms of LMIs in this section. In Section IV, a numerical example is given to demonstrate the feasibility and effectiveness of the proposed approach. Finally, the conclusions are given in Section V.

Notation: The notations in this paper are quite standard. \mathbb{R}^n and $\mathbb{R}^{n \times m}$ denote, respectively, the n -dimensional Euclidean space and the set of all $n \times m$ real matrices. A^T denotes the transpose of matrix A . $X \leq Y$ or $X < Y$, respectively, where X and Y are symmetric matrices, means that $X - Y$ is negative semi-definite or negative definite, respectively. I is the identity matrix with a compatible dimension (without confusion). $diag[q_1, \dots, q_n]$ represents a block-diagonal, $q_1 \cdots q_n$ as the diagonal elements. \star denotes a symmetric matrix, where \star represents the entries implied by symmetry. The matrices, if not explicitly stated, are assumed to have compatible dimensions. $!$ denotes factorial for combinatoric expression. Let $K(h)$ be the set of r -tuples defined as [19]:

$$K(h) = \{ (k_1 k_2 \cdots k_r) : k_1 + k_2 + \cdots + k_r = h, \\ \forall k_i \in I^+ (\text{positive integers}), i = 1, 2, \dots, r \}$$

where h is the total polynomial degree. Since the number of

fuzzy base is r , the number of elements in $K(h)$ is expressed by $J(h) = (r + h - 1)! / (h!(r - 1)!)$. For example, $r = 2$, $h = 3$

$$\begin{aligned} J(3) &= (2 + 3 - 1)! / (3!(2 - 1)!) = 4 \\ K(3) &= \{ (30), (21), (12), (03) \} \\ &= \{ t(1), t(2), t(3), t(4) \} \end{aligned}$$

For clarity, the following notations are adopted:

$$\begin{aligned} k &= k_1 k_2 \cdots k_r \\ \mu^k &= \mu_1^{k_1} \mu_2^{k_2} \cdots \mu_r^{k_r} \\ e_i &= 0 \cdots \underbrace{1}_{i^{th}} \cdots 0 \\ k - e_i &= k_1 k_2 \cdots (k_i - 1) \cdots k_r \\ \pi(k) &= (k_1!) (k_2!) \cdots (k_r!). \end{aligned}$$

II. PRELIMINARIES

To begin, consider the following T-S fuzzy neutral system:

$$\begin{aligned} \dot{x}(t) &= \sum_{i=1}^r \mu_i(t) [A_i x(t) + A_{ci} x(t - h(t)) + A_{di} \dot{x}(t - h(t)) \\ &\quad + B_i u(t)] \\ &= A(t)x(t) + A_c(t)x(t - h(t)) + A_d(t)\dot{x}(t - h(t)) \\ &\quad + B(t)u(t) \\ x(t) &= \phi(t), \quad \forall t \in [-\max\{h_M\}, 0], \quad i = 1, \dots, r \end{aligned} \quad (1)$$

where r is the number of fuzzy rules. The state of system $x(t) \in \mathbb{R}^{n \times 1}$ and the input $u(t) \in \mathbb{R}^{m \times 1}$. The matrices A_i , A_{ci} , $A_{di} \in \mathbb{R}^{n \times n}$, $B_i \in \mathbb{R}^{n \times m}$ are system matrices, and initial vector $\phi(t)$ belongs to the set of continuous functions. The time-varying delay $h(t)$ satisfies $0 \leq h(t) \leq h_M$ and $0 \leq \dot{h}(t) \leq h_D$. $\mu_i(t) = \omega_i(t) / \sum_{i=1}^r \omega_i(t)$, $\omega_i(t) = \prod_{j=1}^p M_{ij}(\xi(t))$. M_{ij} is the membership degree of $\xi(t)$, and $\omega_i(t) \geq 0$ for all t , $i = 1, \dots, r$. It is clear that $\mu_i(t) \geq 0$, and $\sum_{i=1}^r \mu_i(t) = 1$.

The state feedback fuzzy controller for T-S fuzzy neutral system (1) is represented as follows.

$$\begin{aligned} u(t) &= \sum_{k \in K(r-1), r \geq 2} \mu^k F_k x(t) \\ &= F(t)x(t) \end{aligned} \quad (2)$$

By substituting (2) into (1), the closed-loop system can be obtained as (3).

$$\begin{aligned} \dot{x}(t) &= (A(t) + B(t)F(t))x(t) + A_c(t)x(t - h(t)) \\ &\quad + A_d(t)\dot{x}(t - h(t)) \end{aligned} \quad (3)$$

III. MAIN RESULTS

Before discussing the proof of the theorems, here are some lemmas which are used in the proof.

Lemma 1: [20] For any positive symmetric constant matrix $R_1 \in \mathbb{R}^{n \times n}$ and a scalar $h_M > 0$, if there exists a vector function $\dot{x}(s) : [0, h_M] \rightarrow \mathbb{R}^n$ such that the integrals $\int_{t-h(t)}^t \dot{x}(s) R_1 \dot{x}(s) ds$ and $\int_{t-h(t)}^t \dot{x}(s) ds$ are well defined, then the following inequality holds:

$$-h_M \int_{t-h_M}^t \dot{x}^T(s) R_1 \dot{x}(s) ds \leq -h(t) \int_{t-h(t)}^t \dot{x}^T(s) R_1 \dot{x}(s) ds$$

$$\leq - \left(\int_{t-h(t)}^t \dot{x}(s) ds \right)^T R_1 \left(\int_{t-h(t)}^t \dot{x}(s) ds \right).$$

Lemma 2: [21] (Pólya's theorem) For a positive integer r , $\{\Delta_r: (\mu_1, \dots, \mu_r) \mid \mu_i \geq 0, \sum_{i=1}^r \mu_i = 1\}$. If a real homogeneous polynomial $F(\mu_1, \dots, \mu_r)$ is positive definite, then for a sufficiently large d , all the coefficients of

$$(\mu_1 + \cdots + \mu_r)^d F(\mu_1, \dots, \mu_r)$$

are positive.

Lemma 3: [16] Consider the T-S fuzzy system,

$$\dot{x}(t) = \sum_{i=1}^r \mu_i(t) [A_i x(t) + B_i u(t)] \quad (4)$$

where $x(t)$ is the state of system, $u(t)$ is the control input, $A(t) = \sum_{i=1}^r \mu_i(\xi(t)) A_i$, $B(t) = \sum_{i=1}^r \mu_i(\xi(t)) B_i$. The T-S fuzzy system (4) is asymptotically stabilizable via the state feedback controller $G_k = \bar{G}_k X^{-1}$ if and only if there exist a symmetric positive definite matrix $X > 0$ and $d \in \mathbb{N}$ such that:

$$\begin{aligned} \sum_{k' \in K(d)} \sum_{i=1}^r \frac{d!}{\pi(k')} \left(\frac{(k_i - k'_i)}{\pi(k - k')} (A_i X + \star) \right. \\ \left. + (B_i G_{k-k'-e_i} + \star) \right) < 0 \end{aligned} \quad (5)$$

where $k \in K(d + r)$, $k \succ k'$, symbol \succ is the componentwise.

Before discussing the following stability conditions, we first define the following function:

$$\begin{aligned} P_1(t) &= \sum_{i=1}^r \mu_i(t) P_{1i}, \quad P_2(t) = \sum_{i=1}^r \mu_i(t) P_{2i} \\ P_3(t) &= \sum_{i=1}^r \mu_i(t) P_{3i}, \quad S_1(t) = \sum_{i=1}^r \mu_i(t) S_{1i} \\ P_4(t) &= \sum_{j=1}^r \mu_j(t) P_{4j}. \end{aligned}$$

The main result on the asymptotic stability of the T-S fuzzy neutral system (1) is propounded in the following theorem.

Theorem 1: If there exist integers $d_a > 0$, $d_b > 0$, some positive matrices $\bar{P}_0, \bar{P}_{1i}, \bar{P}_{2i}, P_{3i}, P_{4j} \in \mathbb{R}^{n \times n}$, and matrix $\bar{S}_{1i}^T = [\bar{S}_{11i}^T, \bar{S}_{12i}^T, \bar{S}_{13i}^T, \bar{S}_{14i}^T], \bar{S}_{11i}^T, \dots, \bar{S}_{14i}^T \in \mathbb{R}^{n \times n}$ such that the following inequalities (6) and (7) are satisfied for some positive scalar g_1, g_2, g_3 and $0 \leq h(t) \leq h_M$, $0 \leq \dot{h}(t) \leq h_D$ then the fuzzy neutral system (1) is asymptotically stabilizable via the state feedback controller $F_k = \bar{F}_k X^{-1}$, $X \in \mathbb{R}^{n \times n}$.

$$\Omega < 0 \quad (6)$$

$$P_{3i} < P_{4j} \quad (7)$$

where

$$\Omega = \text{diag} \left[\Omega^{t_a(1), t_b(1)}, \dots, \Omega^{t_a(J(d+2)), t_b(J(d+1))} \right]$$

$$\Omega^{k_a, k_b} = \sum_{k'_a \in K(d_a)} \frac{d_a!}{\pi(k'_a)} \sum_{k'_b \in K(d_b)} \frac{d_b!}{\pi(k'_b)} \bar{\Omega}$$

$$k_a \in K(d_a + 2), \quad k_b \in K(d_b + 1), \quad k_a \succ k'_a, \quad k_b \succ k'_b$$

$$\bar{\Omega} = \begin{bmatrix} \bar{\Omega}_{11} & \bar{\Omega}_{12} & \bar{\Omega}_{13} & \bar{\Omega}_{14} & \bar{\Omega}_{15} \\ \star & \bar{\Omega}_{22} & \bar{\Omega}_{23} & \bar{\Omega}_{24} & \bar{\Omega}_{25} \\ \star & \star & \bar{\Omega}_{33} & \bar{\Omega}_{34} & \bar{\Omega}_{35} \\ \star & \star & \star & \bar{\Omega}_{44} & \bar{\Omega}_{45} \\ \star & \star & \star & \star & \bar{\Omega}_{55} \end{bmatrix}$$

$$\bar{\Omega}_{11} = \sum_{i=1}^r \frac{(r-1)!(k_{ai} - k'_{ai})}{\pi(k_a - k'_a)} \frac{1!}{\pi(k_b - k'_b)} [(\bar{S}_{11i} + \star) + \bar{P}_{1i}]$$

$$+ (A_i X + \star) + \sum_{i=1}^r \frac{1!}{\pi(k_b - k'_b)} [(B_i \bar{F}_{k_a - k'_a - e_i} + \star)]$$

$$\bar{\Omega}_{12} = \sum_{i=1}^r \frac{(r-1)!(k_{ai} - k'_{ai})}{\pi(k_a - k'_a)} \frac{1!}{\pi(k_b - k'_b)} [g_1 X^T A_i^T$$

$$+ \bar{S}_{12i}^T + A_{ci} X - \bar{S}_{11i}^T]$$

$$+ \sum_{i=1}^r \frac{1!}{\pi(k_b - k'_b)} [g_1 \bar{F}_{k_a - k'_a - e_i}^T B_i^T]$$

$$\bar{\Omega}_{13} = \sum_{i=1}^r \frac{(r-1)!(k_{ai} - k'_{ai})}{\pi(k_a - k'_a)} \frac{1!}{\pi(k_b - k'_b)} [g_2 X^T A_i^T$$

$$+ \bar{S}_{13i}^T] + \sum_{i=1}^r \frac{1!}{\pi(k_b - k'_b)} [g_2 \bar{F}_{k_a - k'_a - e_i}^T B_i^T]$$

$$+ \sum_{i=1}^r \frac{r!}{\pi(k_a - k'_a)} \frac{1!}{\pi(k_b - k'_b)} [\bar{P}_0^T - X]$$

$$\bar{\Omega}_{14} = \sum_{i=1}^r \frac{(r-1)!(k_{ai} - k'_{ai})}{\pi(k_a - k'_a)} \frac{1!}{\pi(k_b - k'_b)} [A_{di} X + \bar{S}_{14i}^T$$

$$+ g_3 X^T A_i^T] + \sum_{i=1}^r \frac{1!}{\pi(k_b - k'_b)} [g_3 \bar{F}_{k_a - k'_a - e_i}^T B_i^T]$$

$$\bar{\Omega}_{15} = \sum_{i=1}^r \frac{(r-1)!(k_{ai} - k'_{ai})}{\pi(k_a - k'_a)} \frac{1!}{\pi(k_b - k'_b)} [\bar{S}_{11i}]$$

$$\bar{\Omega}_{22} = \sum_{i=1}^r \frac{(r-1)!(k_{ai} - k'_{ai})}{\pi(k_a - k'_a)} \frac{1!}{\pi(k_b - k'_b)} [(g_1 A_{ci} X + \star)$$

$$- (\bar{S}_{12i} + \star)]$$

$$+ \sum_{j=1}^r \frac{r!}{\pi(k_a - k'_a)} [-(1 - h_D) \bar{P}_{1j}]$$

$$\bar{\Omega}_{23} = \sum_{i=1}^r \frac{(r-1)!(k_{ai} - k'_{ai})}{\pi(k_a - k'_a)} \frac{1!}{\pi(k_b - k'_b)} [g_2 X^T A_{ci}^T$$

$$- \bar{S}_{13i}^T] + \sum_{i=1}^r \frac{r!}{\pi(k_a - k'_a)} \frac{1!}{\pi(k_b - k'_b)} [-g_1 X]$$

$$\bar{\Omega}_{24} = \sum_{i=1}^r \frac{(r-1)!(k_{ai} - k'_{ai})}{\pi(k_a - k'_a)} \frac{1!}{\pi(k_b - k'_b)} [g_3 X^T A_{ci}^T$$

$$+ g_1 A_{di} X - \bar{S}_{14i}^T]$$

$$\bar{\Omega}_{25} = \sum_{i=1}^r \frac{(r-1)!(k_{ai} - k'_{ai})}{\pi(k_a - k'_a)} \frac{1!}{\pi(k_b - k'_b)} [\bar{S}_{12i}]$$

$$\bar{\Omega}_{33} = \sum_{i=1}^r \frac{(r-1)!(k_{ai} - k'_{ai})}{\pi(k_a - k'_a)} \frac{1!}{\pi(k_b - k'_b)} [\bar{P}_{2i} + h_M \bar{P}_{3i}]$$

$$+ \sum_{i=1}^r \frac{r!}{\pi(k_a - k'_a)} \frac{1!}{\pi(k_b - k'_b)} [(-g_2 X + \star)]$$

$$\bar{\Omega}_{34} = \sum_{i=1}^r \frac{(r-1)!(k_{ai} - k'_{ai})}{\pi(k_a - k'_a)} \frac{1!}{\pi(k_b - k'_b)} [g_2 A_{di} X]$$

$$+ \sum_{i=1}^r \frac{r!}{\pi(k_a - k'_a)} \frac{1!}{\pi(k_b - k'_b)} [-g_3 X]$$

$$\bar{\Omega}_{35} = \sum_{i=1}^r \frac{(r-1)!(k_{ai} - k'_{ai})}{\pi(k_a - k'_a)} \frac{1!}{\pi(k_b - k'_b)} [\bar{S}_{13i}]$$

$$\bar{\Omega}_{44} = \sum_{i=1}^r \frac{(r-1)!(k_{ai} - k'_{ai})}{\pi(k_a - k'_a)} \frac{1!}{\pi(k_b - k'_b)} [(g_3 A_{di} X + \star)]$$

$$+ \sum_{j=1}^r \frac{r!}{\pi(k_a - k'_a)} [-(1 - h_D) \bar{P}_{2j}]$$

$$\bar{\Omega}_{45} = \sum_{i=1}^r \frac{(r-1)!(k_{ai} - k'_{ai})}{\pi(k_a - k'_a)} \frac{1!}{\pi(k_b - k'_b)} [\bar{S}_{14i}]$$

$$\bar{\Omega}_{55} = \sum_{i=1}^r \frac{(r-1)!(k_{ai} - k'_{ai})}{\pi(k_a - k'_a)} \frac{1!}{\pi(k_b - k'_b)} [-h_M^{-1} \bar{P}_{3i}].$$

Proof: Firstly, let us consider the Lyapunov-Krasovskii function

$$V(t) = V_1(t) + V_2(t) + V_3(t) + V_4(t) \quad (8)$$

where

$$V_1(t) = x^T(t) P_0 x(t)$$

$$V_2(t) = \int_{t-h(t)}^t x^T(s) P_1(s) x(s) ds$$

$$V_3(t) = \int_{t-h(t)}^t \dot{x}^T(s) P_2(s) \dot{x}(s) ds$$

$$V_4(t) = \int_{t-h_M}^t (s - (t - h_M)) \dot{x}^T(s) P_3(s) \dot{x}(s) ds$$

and $P_0, P_1(s), P_2(s), P_3(s)$ are symmetric positive definite matrices.

The time derivative of Lyapunov-Krasovskii function is

$$\dot{V}(t) = \dot{V}_1(t) + \dot{V}_2(t) + \dot{V}_3(t) + \dot{V}_4(t) \quad (9)$$

where

$$\dot{V}_1(t) = \dot{x}^T(t) P_0 x(t) + x^T(t) P_0 \dot{x}(t)$$

$$\dot{V}_2(t) = x^T(t) P_1(t) x(t)$$

$$- (1 - \dot{h}(t)) x^T(t - h(t)) P_1(t - h(t)) x(t - h(t))$$

$$\dot{V}_3(t) = \dot{x}^T(t) P_2(t) \dot{x}(t)$$

$$- (1 - \dot{h}(t)) \dot{x}^T(t - h(t)) P_2(t - h(t)) \dot{x}(t - h(t))$$

$$\dot{V}_4(t) = h_M \dot{x}^T(t) P_3(t) \dot{x}(t) - \int_{t-h_M}^t \dot{x}^T(s) P_3(s) \dot{x}(s) ds.$$

According to $0 \leq \dot{h}(t) \leq h_D$, we can obtain

$$\dot{V}_2(t) = x^T(t) P_1(t) x(t) - (1 - h_D)$$

$$\times x^T(t - h(t)) P_1(t - h(t)) x(t - h(t)) \quad (10)$$

$$\dot{V}_3(t) = \dot{x}^T(t) P_2(t) \dot{x}(t) - (1 - h_D)$$

$$\times \dot{x}^T(t - h(t)) P_2(t - h(t)) \dot{x}(t - h(t)). \quad (11)$$

From Newton-Leibniz formula and the T-S fuzzy neutral system in (1), the following equalities are always hold:

$$\Pi_1 = 2\xi^T(t)S_1(t)[x(t) - x(t-h(t)) - \int_{t-h(t)}^t \dot{x}(s)ds] = 0 \quad (12)$$

$$\Pi_2 = 2\xi^T(t)S_2(t)[A(t)x(t) + A_c(t)x(t-h(t)) + A_d(t)\dot{x}(t-h(t)) + B(t)K(t)x(t) - \dot{x}(t)] = 0 \quad (13)$$

$$\Pi_3 = h_M \xi^T(t)S_1(t)P_4^{-1}(t)S_1^T(t)\xi(t) - \int_{t-h_M}^t \xi^T(t)S_1(t)P_4^{-1}(t)S_1^T(t)\xi(t)ds = 0 \quad (14)$$

where $\xi^T(t) = [x^T(t) \quad x^T(t-h(t)) \quad \dot{x}^T(t) \quad \dot{x}^T(t-h(t))]$.

From the condition $0 \leq h(t) \leq h_M$, we can obtain the following result

$$\begin{aligned} \Pi_3 &= h_M \xi^T(t)S_1(t)P_4^{-1}(t)S_1^T(t)\xi(t) \\ &\quad - \int_{t-h_M}^t \xi^T(t)S_1(t)P_4^{-1}(t)S_1^T(t)\xi(t)ds \\ &\leq h_M \xi^T(t)S_1(t)P_4^{-1}(t)S_1^T(t)\xi(t) \\ &\quad - \int_{t-h(t)}^t \xi^T(t)S_1(t)P_4^{-1}(t)S_1^T(t)\xi(t)ds. \end{aligned} \quad (15)$$

By (9)-(15) with $P_3(s) \geq P_4(t)$ we can obtain

$$\begin{aligned} \dot{V}(t) &\leq \Lambda(t) + h_M \xi^T(t)S_1(t)P_4^{-1}(t)S_1^T(t)\xi(t) \\ &\quad - \int_{t-h(t)}^t [(\dot{x}^T(s)P_3(s) + \xi^T(t)S_1(t)) \\ &\quad \times P_3^{-1}(s)(P_3(s)\dot{x}(s) + S_1^T(t)\xi(t))]ds \\ &\leq \Lambda(t) + h_M \xi^T(t)S_1(t)P_4^{-1}(t)S_1^T(t)\xi(t) \end{aligned} \quad (16)$$

where

$$\begin{aligned} \Lambda(t) &= \dot{x}^T(t)P_0x(t) + x^T(t)P_0\dot{x}(t) + x^T(t)P_1(t)x(t) \\ &\quad - (1-h_D)x^T(t-h(t))P_1(t-h(t))x(t-h(t)) \\ &\quad - (1-h_D)\dot{x}^T(t-h(t))P_2(t-h(t))\dot{x}(t-h(t)) \\ &\quad + \dot{x}^T(t)P_2(t)\dot{x}(t) + h_M \dot{x}^T(t)P_3(t)\dot{x}(t) \\ &\quad + 2\xi^T(t)S_1(t)[x(t) - x(t-h(t))] \\ &\quad + 2\xi^T(t)S_2(t)[A(t)x(t) + A_c(t)x(t-h(t)) \\ &\quad + A_d(t)\dot{x}(t-h(t)) + B(t)K(t)x(t) - \dot{x}(t)]. \end{aligned}$$

In order for $\dot{V}(x(t)) < 0$ for all $x(t) \neq 0$, (16) should be negative. By Schur complement and pre- and post-multiplying both sides with $\text{diag}[X, X, X, X, X]$ and define $X = S_{21}^{-1}$, $S_{22} = g_1 S_{21}$, $S_{23} = g_2 S_{21}$, $S_{24} = g_3 S_{21}$, we can get the following result from (16)

$$\xi^T(t)\Xi(t)\xi(t) < 0 \quad (17)$$

where

$$\Xi(t) = \begin{bmatrix} \Xi_{11} & \Xi_{12} & \Xi_{13} & \Xi_{14} & \Xi_{15} \\ \star & \Xi_{22} & \Xi_{23} & \Xi_{24} & \Xi_{25} \\ \star & \star & \Xi_{33} & \Xi_{34} & \Xi_{35} \\ \star & \star & \star & \Xi_{44} & \Xi_{45} \\ \star & \star & \star & \star & \Xi_{55} \end{bmatrix}$$

$$\begin{aligned} \Xi_{11} &= [(\bar{S}_{11}(t) + \star) + \bar{P}_1(t) + (A(t)X + \star) + (B(t)\bar{F}(t) + \star)] \\ \Xi_{12} &= [A_c(t)X + g_1 X^T A(t)^T + \bar{S}_{12}^T(t) - \bar{S}_{11}^T(t) \\ &\quad + g_1 \bar{F}^T(t)B^T(t)] \\ \Xi_{13} &= [g_2 X^T A(t)^T + \bar{S}_{13}^T(t) + \bar{P}_0^T - X + g_2 \bar{F}^T(t)B^T(t)] \\ \Xi_{14} &= [A_d(t)X + g_3 X^T A(t)^T + \bar{S}_{14}^T(t) + g_3 \bar{F}^T(t)B^T(t)] \\ \Xi_{15} &= [\bar{S}_{11}(t)] \\ \Xi_{22} &= [(g_1 A_c(t)X + \star) - (\bar{S}_{12}(t) + \star) - (1-h_D)\bar{P}_1(t-h(t))] \\ \Xi_{23} &= [g_2 X^T A_c(t)^T - \bar{S}_{13}^T(t) - g_1 X] \\ \Xi_{24} &= [g_1 A_d(t)X + g_3 X^T A_c(t)^T - \bar{S}_{14}^T(t)] \\ \Xi_{25} &= [\bar{S}_{12}(t)] \\ \Xi_{33} &= [\bar{P}_2(t) + h_M \bar{P}_3(t) - (g_2 X + \star)] \\ \Xi_{34} &= [g_2 A_d(t)X - g_3 X] \\ \Xi_{35} &= [\bar{S}_{13}(t)] \\ \Xi_{44} &= [(g_3 A_d^T(t)X + \star) - (1-h_D)\bar{P}_2(t-h(t))] \\ \Xi_{45} &= [\bar{S}_{14}(t)] \\ \Xi_{55} &= [-h_M^{-1}\bar{P}_3(t)] \\ \bar{S}_{11}(t) &= X S_{11}(t)X, \quad \bar{S}_{12}(t) = X S_{12}(t)X, \quad \bar{P}_0(t) = X P_0(t)X, \\ \bar{S}_{13}(t) &= X S_{13}(t)X, \quad \bar{S}_{14}(t) = X S_{14}(t)X, \quad \bar{P}_1(t) = X P_1(t)X, \\ \bar{P}_2(t) &= X P_2(t)X, \quad \bar{P}_1(t-h(t)) = X P_1(t-h(t))X \\ \bar{P}_3(t) &= X P_3(t)X, \quad \bar{P}_2(t-h(t)) = X P_2(t-h(t))X. \end{aligned}$$

Clearly, (17) is equivalent to (18)

$$\Phi(t) < 0 \quad (18)$$

where

$$\Phi(t) = \begin{bmatrix} \Phi_{11} & \Phi_{12} & \Phi_{13} & \Phi_{14} & \Phi_{15} \\ \star & \Phi_{22} & \Phi_{23} & \Phi_{24} & \Phi_{25} \\ \star & \star & \Phi_{33} & \Phi_{34} & \Phi_{35} \\ \star & \star & \star & \Phi_{44} & \Phi_{45} \\ \star & \star & \star & \star & \Phi_{55} \end{bmatrix}$$

$$\begin{aligned} \Phi_{11} &= \sum_{j=1}^r \mu_j [(B(t)\bar{F}(t) + \star)] + \sum_{j=1}^r \mu_j \left(\sum_{i=1}^r \mu_i \right)^{r-1} [\bar{P}_1(t) \\ &\quad + (\bar{S}_{11}(t) + \star) + (A(t)X + \star)] \end{aligned}$$

$$\begin{aligned} \Phi_{12} &= \sum_{j=1}^r \mu_j \left(\sum_{i=1}^r \mu_i \right)^{r-1} [A_c(t)X + g_1 X^T A(t)^T \\ &\quad + \bar{S}_{12}^T(t) - \bar{S}_{11}^T(t)] + \sum_{j=1}^r \mu_j [g_1 \bar{F}^T(t)B^T(t)] \end{aligned}$$

$$\begin{aligned} \Phi_{13} &= \sum_{j=1}^r \mu_j \left(\sum_{i=1}^r \mu_i \right)^{r-1} [g_2 X^T A(t)^T + \bar{S}_{13}^T(t)] \\ &\quad + \sum_{j=1}^r \mu_j [g_2 \bar{F}^T(t)B^T(t)] + \sum_{j=1}^r \mu_j \left(\sum_{i=1}^r \mu_i \right)^r [\bar{P}_0^T - X] \end{aligned}$$

$$\begin{aligned} \Phi_{14} &= \sum_{j=1}^r \mu_j \left(\sum_{i=1}^r \mu_i \right)^{r-1} [A_d(t)X + g_3 X^T A(t)^T + \bar{S}_{14}^T(t)] \\ &\quad + \sum_{j=1}^r \mu_j [g_3 \bar{F}^T(t)B^T(t)] \end{aligned}$$

$$\begin{aligned}
\Phi_{15} &= \sum_{j=1}^r \mu_j \left(\sum_{i=1}^r \mu_i \right)^{r-1} [\bar{S}_{11}(t)] \\
\Phi_{22} &= \sum_{j=1}^r \mu_j \left(\sum_{i=1}^r \mu_i \right)^{r-1} [(g_1 A_c(t) X + \star) - (\bar{S}_{12}(t) + \star)] \\
&\quad - \left(\sum_{i=1}^r \mu_i \right)^r [(1 - h_D) \bar{P}_1(t - h(t))] \\
\Phi_{23} &= \sum_{j=1}^r \mu_j \left(\sum_{i=1}^r \mu_i \right)^{r-1} [g_2 X^T A_c(t)^T - \bar{S}_{13}^T(t)] \\
&\quad - \sum_{j=1}^r \mu_j \left(\sum_{i=1}^r \mu_i \right)^r [g_1 X] \\
\Phi_{24} &= \sum_{j=1}^r \mu_j \left(\sum_{i=1}^r \mu_i \right)^{r-1} [g_1 A_d(t) X + g_3 X^T A_c(t)^T - \bar{S}_{14}^T(t)] \\
\Phi_{25} &= \sum_{j=1}^r \mu_j \left(\sum_{i=1}^r \mu_i \right)^{r-1} [\bar{S}_{12}(t)] \\
\Phi_{33} &= \sum_{j=1}^r \mu_j \left(\sum_{i=1}^r \mu_i \right)^{r-1} [\bar{P}_2(t) + h_M \bar{P}_3(t)] \\
&\quad - \sum_{j=1}^r \mu_j \left(\sum_{i=1}^r \mu_i \right)^r [(g_2 X + \star)] \\
\Phi_{34} &= \sum_{j=1}^r \mu_j \left(\sum_{i=1}^r \mu_i \right)^{r-1} [g_2 A_d(t) X] \\
&\quad - \sum_{j=1}^r \mu_j \left(\sum_{i=1}^r \mu_i \right)^r [g_3 X] \\
\Phi_{35} &= \sum_{j=1}^r \mu_j \left(\sum_{i=1}^r \mu_i \right)^{r-1} [\bar{S}_{13}(t)] \\
\Phi_{44} &= \sum_{j=1}^r \mu_j \left(\sum_{i=1}^r \mu_i \right)^{r-1} [(g_3 A_d(t) X + \star)] \\
&\quad + \left(\sum_{i=1}^r \mu_i \right)^r [(1 - h_D) \bar{P}_2(t - h(t))] \\
\Phi_{45} &= \sum_{j=1}^r \mu_j \left(\sum_{i=1}^r \mu_i \right)^{r-1} [\bar{S}_{14}(t)] \\
\Phi_{55} &= \sum_{j=1}^r \mu_j \left(\sum_{i=1}^r \mu_i \right)^{r-1} [-h_M^{-1} \bar{P}_3(t)].
\end{aligned}$$

By applying Lemma 2 and Lemma 3 to (18), yields

$$\begin{aligned}
&\left(\sum_{i=1}^r \mu_i(t) \right)^{d_a} \left(\sum_{j=1}^r \mu_j(t - h(t)) \right)^{d_b} \Phi(t) \\
&= \sum_{k'_a \in K(d_a)} \frac{d_a!}{\pi(k'_a)} \sum_{k'_b \in K(d_b)} \frac{d_b!}{\pi(k'_b)} \bar{\Omega} < 0 \\
&= \sum_{k'_a \in K(d_a)} \sum_{k'_b \in K(d_b)} \Omega^{k_a, k_b} < 0.
\end{aligned} \tag{19}$$

Therefore, if (19) is satisfied which implies that $\Omega < 0$, and the closed-loop T-S fuzzy neutral system is asymptotically stable. This completes the proof of the theorem. \blacksquare

IV. NUMERICAL EXAMPLE

In this section, a numerical example is provided to demonstrate the validity and feasibility of the proposed result.

Example 1: Consider the following T-S fuzzy neutral system:

$$\begin{aligned}
\dot{x}(t) &= \sum_{i=1}^2 \mu_i(t) [A_i x(t) + A_{ci} x(t - h(t)) + A_{di} \dot{x}(t - h(t)) \\
&\quad + B_i u(t)]
\end{aligned} \tag{20}$$

where

$$\begin{aligned}
A_1 &= \begin{bmatrix} 0.3 & 0.6 \\ 0.8 & 1 \end{bmatrix}, A_2 = \begin{bmatrix} 1 & 0.3 \\ 1 & 0.6 \end{bmatrix}, \\
A_{c1} &= \begin{bmatrix} 0.5 & 0.9 \\ 0 & 2 \end{bmatrix}, A_{c2} = \begin{bmatrix} 0.9 & 0 \\ 1 & 1.6 \end{bmatrix}, \\
A_{d1} &= \begin{bmatrix} -0.5 & 1 \\ 0.4 & 0.3 \end{bmatrix}, A_{d2} = \begin{bmatrix} 0.3 & 0 \\ 0.7 & -0.2 \end{bmatrix}, \\
B_1 = B_2 &= \begin{bmatrix} 1 \\ 1 \end{bmatrix}, h(t) = 0.3 + 0.2 \cos(t).
\end{aligned}$$

By applying the convex optimization problem in Theorem 1 with $g_1 = 0.32$, $g_2 = 2.55$, $g_3 = 0.1$, $h_M = 0.5$, and $h_D = 0.2$, the following matrices and controller gain can be obtained:

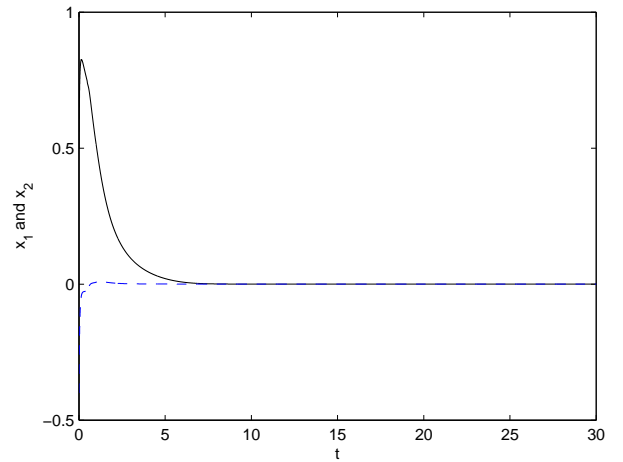


Fig. 1. The state response for closed-loop T-S fuzzy neutral systems with initial condition $x(0) = [0.5, -0.4]$.

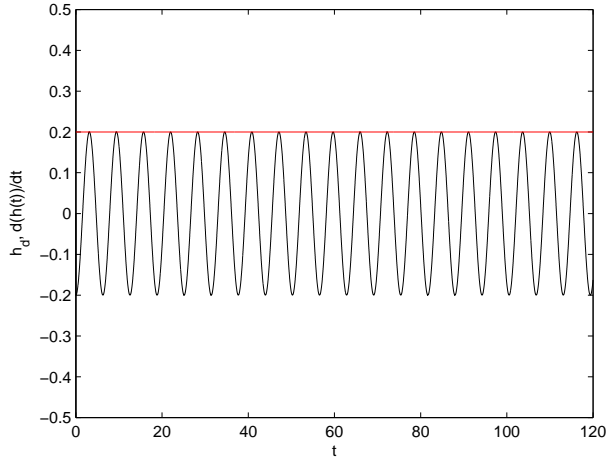


Fig. 2. The variation of $\dot{h}(t)$ and h_D .

$$\begin{aligned}
 P_0 &= \begin{bmatrix} 14.1414 & 13.7123 \\ 13.7123 & 13.5008 \end{bmatrix}, \quad X = \begin{bmatrix} 0.4453 & 0.2166 \\ 0.2166 & 0.2209 \end{bmatrix}, \\
 P_{11} &= \begin{bmatrix} 8.9149 & 8.3762 \\ 8.3762 & 8.1529 \end{bmatrix}, \quad P_{12} = \begin{bmatrix} 7.4549 & 7.0032 \\ 7.0032 & 6.9467 \end{bmatrix}, \\
 P_{21} &= \begin{bmatrix} 0.8856 & 0.3300 \\ 0.3300 & 0.2413 \end{bmatrix}, \quad P_{22} = \begin{bmatrix} 1.4665 & 0.7449 \\ 0.7449 & 0.5225 \end{bmatrix}, \\
 P_{31} &= \begin{bmatrix} 0.7179 & 0.1753 \\ 0.1753 & 0.1872 \end{bmatrix}, \quad P_{32} = \begin{bmatrix} 1.1204 & 0.1160 \\ 0.1160 & 0.0456 \end{bmatrix}, \\
 F_{10} &= [-1.3569 \quad -23.2738], \quad F_{01} = [-2.2770 \quad -22.3444].
 \end{aligned}$$

The state responses for closed-loop T-S fuzzy neutral system with delay time $0.3 + 0.2\cos(t)$ and $x(0) = [0.5, -0.4]$ is shown in Fig. 1. From the simulation results, it can be seen the designed fuzzy controller ensures the asymptotic stability of the closed-loop T-S fuzzy neutral system. One can observe that the states converge to the equilibrium states after some transient times. Fig. 2 shows the variation of $\dot{h}(t)$ and h_D .

V. CONCLUSIONS

In this paper, a stabilization problem for T-S fuzzy neutral system is investigated. Based on the polynomial technique and some variable transformation, a delay-dependent stabilization condition is proposed for T-S fuzzy neutral system. Furthermore, the results can be formulated in terms of LMI forms. A numerical example is given to illustrate the effectiveness of the proposed methods.

ACKNOWLEDGMENT

This work was supported by the National Science Council of Taiwan, R.O.C., under Grant MOST-103-2221-E-027-089-, and King's College London and the State Scholarship Fund of China Scholarship Council.

REFERENCES

[1] C.-Y. Wang, T.-F. Lee, C.-H. Fang, and J.-H. Chou, "Fuzzy logic-based prognostic score for outcome prediction in esophageal cancer," *IEEE Trans. Inf. Technol. Biomed.*, vol. 16, no. 6, pp. 1224–1230, Aug. 2012.

[2] N. U. M. and J. Khastoo, "Fuzzy logic-based efficiency optimization and high dynamic performance of IPMSM drive system in both transient and steady-state conditions," *IEEE Trans. Ind. Appl.*, vol. 50, no. 6, pp. 4251–4259, Apr. 2014.

[3] S.-H. Tsai, "An improved fuzzy modeling method for a class of multi-input non-affine nonlinear systems," *Optimization Theory and Applications*, vol. 157, no. 1, pp. 287–296, Apr. 2013.

[4] T. Tanaka and H. O. Wang, *Fuzzy Control Systems Design and Analysis: A Linear Matrix Inequality Approach*. New York: Wiley, 2001.

[5] S. Derakhshan, A. Fatehi, and M. G. Sharabiany, "Nonmonotonic observer-based fuzzy controller designs for discrete time T-S fuzzy systems via LMI," *IEEE Trans. Cybernetics.*, vol. 44, no. 12, pp. 2557–2567, Apr. 2014.

[6] K. Tanaka, M. Tanaka, H. Ohtake, and H. O. Wang, "Shared nonlinear control in wireless-based remote stabilization: A theoretical approach," *IEEE/ASME Trans. Mechatronics*, vol. 17, no. 3, pp. 443–453, Mar. 2012.

[7] Y. J. Chen, H. Ohtake, K. Tanaka, and W. J. Wang, "Relaxed stabilization criterion for T-S fuzzy systems by minimum-type piecewise-Lyapunov-function-based switching fuzzy controller," *IEEE Trans. Fuzzy Syst.*, vol. 20, no. 6, pp. 1166–1173, Apr. 2012.

[8] C. Lien and K. Yu, "Robust control for Takagi-Sugeno fuzzy systems with time-varying state and input delays," *Chaos, Solitons & Fractals*, vol. 35, no. 5, pp. 1003–1008, Mar. 2008.

[9] Y. Ariba, F. Gouaisbaut, S. Rahme, and Y. Labit, "Traffic monitoring in transmission control protocol/active queue management networks through a time-delay observer," *IET Control Theory Appl.*, vol. 6, no. 4, pp. 506–517, Mar. 2012.

[10] C. Han, L. Wu, P. Shi, and Q. Zeng, "Passivity and passification of T-S fuzzy descriptor systems with stochastic perturbation and time delay," *IET Control Theory Appl.*, vol. 7, no. 13, pp. 1711–1724, Sep. 2013.

[11] Z. Lin, H. Gao, and H. R. Karimi, "Robust stability and stabilization of uncertain T-S fuzzy systems with time-varying delay an input-output approach," *IEEE Trans. Fuzzy Syst.*, vol. 21, no. 5, pp. 883–897, Dec. 2012.

[12] H. Chen, "Delay-dependent robust H_∞ filter design for uncertain neutral stochastic system with time-varying delay," *IET Signal Processing*, vol. 7, no. 5, pp. 368–381, Jul. 2013.

[13] Y. E. Wang, J. Zhao, and B. Jiang, "Stabilization of a class of switched linear neutral systems under asynchronous switching," *IEEE Trans. Autom. Control*, vol. 58, no. 8, pp. 2114–2119, Mar. 2013.

[14] X. Nian, X. Wang, Y. Wang, and Z. Sun, "Delay-dependent stability for fuzzy neutral system via state matrix decomposition," *IET Control Theory & Applications*, vol. 6, no. 11, pp. 1745–1751, Jul. 2012.

[15] J. Yang, S. Zhong, and L. Xiong, "A descriptor system approach to non-fragile H_∞ control for uncertain fuzzy neutral systems," *Fuzzy Sets & Syst.*, vol. 160, no. 4, pp. 423–438, Jun. 2009.

[16] V. Montagner, R. C. L. F. Oliveira, P. L. D. Peres, and P. A. Bliman, "Linear matrix inequality characterisation for H_∞ and H_2 guaranteed cost gain-scheduling quadratic stabilisation of linear time-varying polytopic systems," *IET Control Theory. Appl.*, vol. 1, no. 6, pp. 1726–1735, Nov. 2007.

[17] H. Zhang, X. Xie, and S. Tong, "Homogenous polynomially parameter-dependent H_∞ filter designs of discrete-time fuzzy systems," *IEEE Trans. Syst. Man Cybern. Part B-Cybern.*, vol. 41, no. 5, pp. 1313–1322, Oct. 2011.

[18] S.-H. Tsai, C. Sun, J. C. Lo, and H. K. Lam, "Relaxed stabilization of t-s fuzzy systems with time-delay," in *International Conference on Fuzzy Systems 2013 (FUZZ-IEEE 2013)*, Hyderabad, India, 2013, pp. 1–6.

[19] J. C. Lo and J. R. Wan, "Studies on linear matrix inequality relaxations for fuzzy control systems via homogeneous polynomials," *IET Control Theory & Appl.*, vol. 4, no. 11, pp. 2293–2302, Jan. 2010.

[20] K. Gu, V. Kharitonov, and J. Chen, *Stability of Time-Delay Systems*. Boston: Birkhauser, 2003.

[21] G. H. Hardy, J. E. Littlewood, and G. Pólya's, *Inequalities*. Cambridge: Cambridge University Press, 1952.